



sysCLOCK PLL Design and User Guide for Nexus Platform

Technical Note

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Acronyms in This Document

A list of acronyms used in this document.

| Acronym | Definition |
|---------|---------------------------------|
| CIB | Common Interface Block |
| DCC | Dynamic Clock Control |
| DCS | Dynamic Clock Select |
| DDR | Double Data Rate |
| DLL | Delay Locked Loop |
| DTR | Digital Temperature Readout |
| GSR | Global Set Reset |
| LMMI | Lattice Memory Mapped Interface |
| MIB | Memory Interface Block |
| PLC | Programmable Logic Cell |
| PLL | Phase Locked Loop |
| SED | Soft Error Detect |

1. Introduction

This usage guide describes the clock resources available in the Lattice Nexus™ Platform architecture, which includes CrossLink™-NX, Certus™-NX, CertusPro™-NX, and MachXO5™-NX product families.

The details are provided for Primary Clocks, Edge Clocks, PLLs, the Internal Oscillator, and clocking elements such as Clock Dividers, Clock Multiplexers, and Clock Stop Blocks available in the Nexus device.

The number of PLLs, Edge Clocks, and Clock Dividers for each device is listed in [Table 1.1](#).

Table 1.1. Number of PLLs, Edge Clocks, and Clock Dividers

| Parameter | Description | LIFCL-17 LFD2NX-17 | LIFCL-33 | LIFCL-40 LFD2NX-40 | LFCPNX-50 | LFCPNX-100 | LFMXO5 |
|----------------------------------|-----------------------------------------------------------------------|-----------------------|----------|-----------------------|-----------|------------|--------|
| Number of PLLs | General purpose Phase Locked Loops. | 2 | 1 | 3 | 3 | 4 | 2 |
| Number of Edge Clocks | Edge Clocks for high-speed interfaces. | 12 | 12 | 12 | 12 | 12 | 8 |
| Number of Edge Clock Dividers | Edge Clock Dividers for high-speed interfaces. | 12 | 12 | 12 | 12 | 12 | 8 |
| Number of Primary Clock Dividers | Programmable Primary Clock dividers for domain crossing applications. | 1 | 1 | 1 | 2 | 2 | 1 |
| Number of DDRDLLs | DDRDLL used for DDR memory and High Speed I/O interfaces | 2 | 2 | 2 | 2 | 2 | 2 |

It is very important to validate the device pinout using the Lattice Radiant™ tool to avoid implementation issues.

2. Clock/Control Distribution Network

Nexus devices provide global clock distribution in the form of global primary clocks. The device is organized into clock regions; each clock region can accommodate 16 primary clocks. For CrossLink-NX, Certus-NX, and MachXO5-NX, there are two clock regions and for CertusPro-NX there are four clock regions. There is a maximum of 64 unique clock input sources. The Nexus primary clocking structure is Edge Clock rich and contains generous low-skew Primary clock resources.

3. Nexus Top-Level View

A top-level view of the major clocking resources for the CrossLink-NX and Certus-NX devices are shown in Figure 3.1. The shaded blocks (PCIe®, upper left PLL, and I/O Bank 2/Bank 6/Bank 7) are not available in the LIFCL-17 and LFD2NX-17 devices. The MIPI_DPHY0 and MIPI_DPHY1 on the top are only available for the CrossLink-NX family.

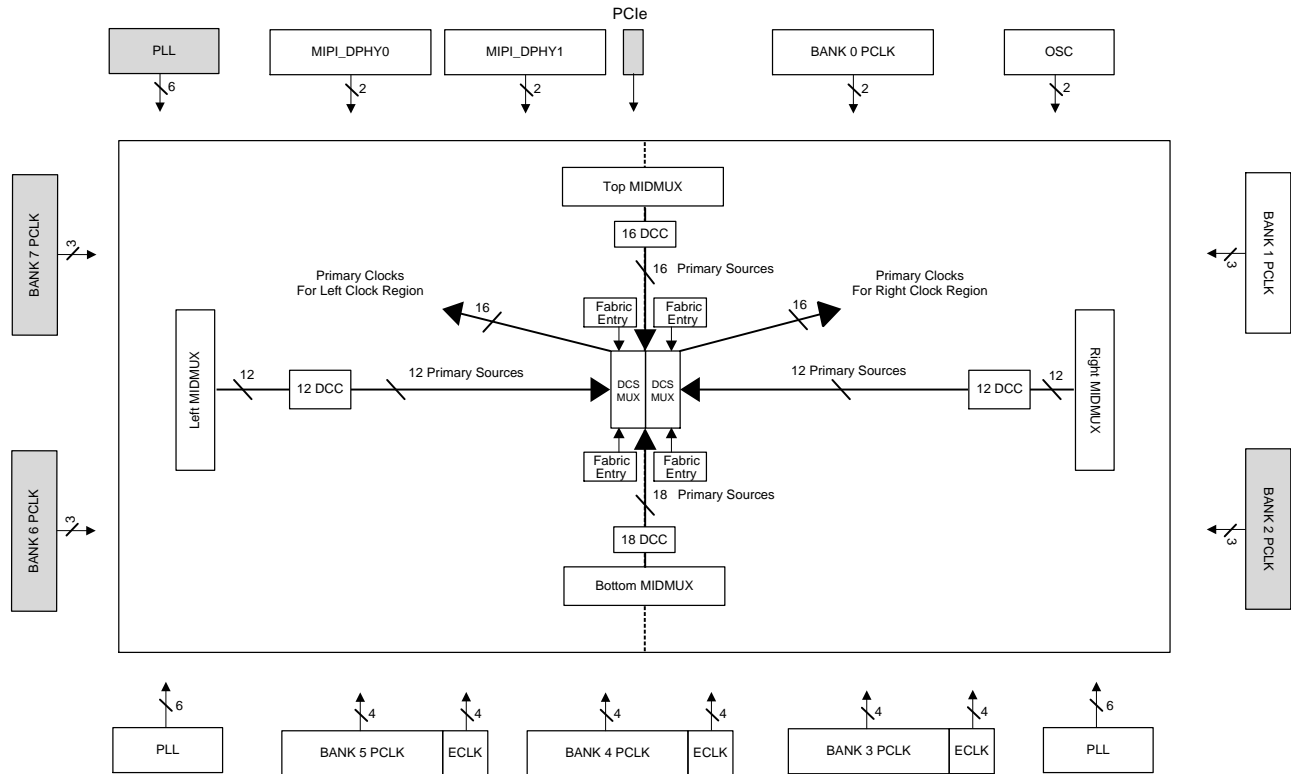
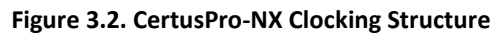


Figure 3.1. CrossLink-NX and Certus-NX Clocking Structure

A top-level view of the major clocking resources for the CertusPro-NX devices are shown in Figure 3.2. The Upper Right PLL is only for LFCPNX-100.



The diagram illustrates the clock distribution architecture. It shows a central processing block with various internal components and external connections. The inputs to the central block are PLL (6), BANK 0 PCLK (2), BANK 1 PCLK (2), and OSC (2). The outputs are BANK 2 PCLK (3), BANK 3 PCLK (2), BANK 4 PCLK (2), BANK 5 PCLK (4), BANK 6 PCLK (4), BANK 7 PCLK (2), BANK 8 PCLK (2), and BANK 9 PCLK (3). The central block contains a TMD block connected to a 16DCC block, which in turn connects to two Fabric Entry blocks. These Fabric Entry blocks connect to two DCS MUX blocks. The DCS MUX blocks are connected to two 12DCC blocks, which then connect to two 18DCC blocks. The 18DCC block is connected to a BMID block. The 12DCC blocks are also connected to LMD and RMD blocks. The diagram also shows connections to various other blocks like PLL, BANK 0 PCLK, BANK 1 PCLK, OSC, LMD, RMD, and various other PCLK and ECLK blocks.

Figure 3.3. MachXO5-NX Clocking Structure

A top-level view of the major clocking resources for the CrossLink-NX-33 device is shown in Figure 3.4.

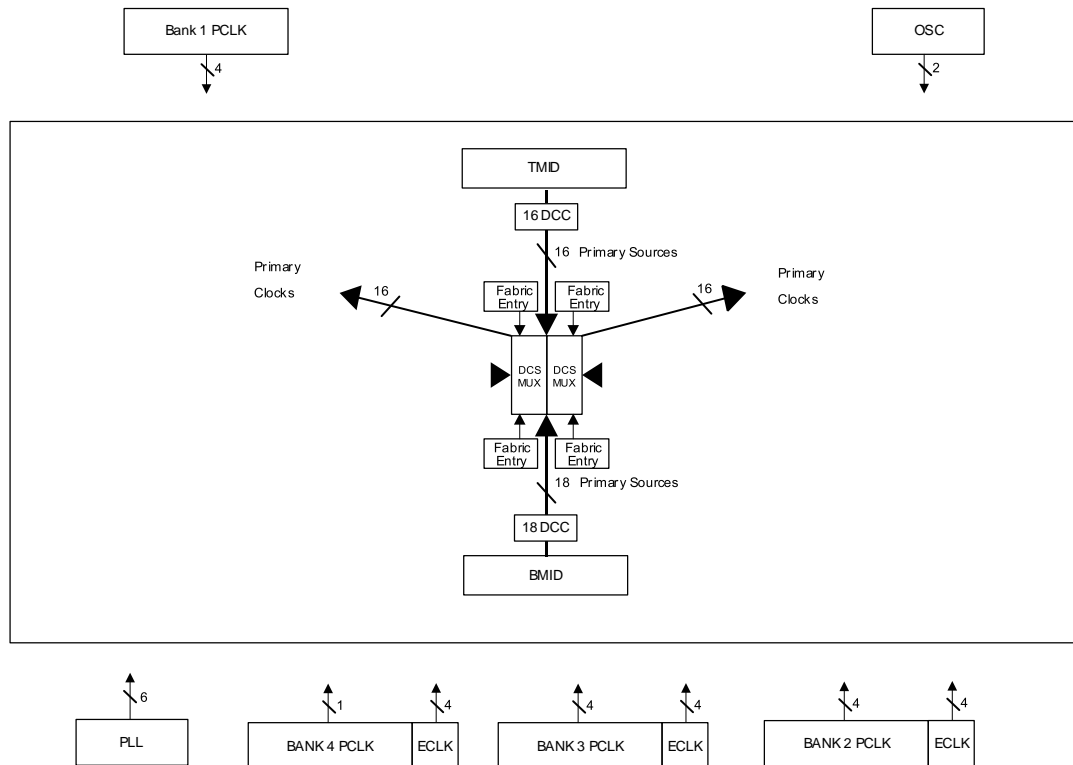


Figure 3.4. CrossLink-NX-33 Clocking Structure

4. Clocking Architecture Overview

This section provides a brief overview of the clocking structure, elements, and PLL. Greater detail is provided starting with the [Appendix A. Primary Clock Sources and Distribution](#) and [Appendix B Pinout Rules for Clocking in Nexus Devices](#) section.

4.1. Primary Clock Network

Up to 32 primary clocks (for CrossLink-NX, Certus-NX, and MachXO5-NX) or 64 primary clocks (for CertusPro-NX) can be selected from up to 64 Primary Clock Sources (PLLs, External Inputs, SerDes, and others) and routed to the Primary Clock Network.

The Primary Clock Network provides low-skew, high fan-out clock distribution to all synchronous elements in the FPGA fabric. The Primary Clock Network is divided into two clocking regions (for CrossLink-NX, Certus-NX, and MachXO5-NX) or four clocking regions (for CertusPro-NX), each region associated with a DCS_CMUX. Each of these regions has 16 clocks that can be distributed to the fabric in the region. Initially, the Lattice Radiant software automatically routes each clock region; up to a maximum of 16 clocks. You can change how the clocks are routed by specifying a preference in the Lattice Radiant project constraints file to locate the clock to specific region.

4.2. Edge Clock Network

Edge Clocks are low skew, high speed clock resources used to clock data into/out of the I/O logic of Nexus devices. There are four Edge Clocks per bank located on the bottom side of the device.

5. Overview of Clocking Components

5.1. Edge Clock Dividers (ECLKDIV)

Edge Clock dividers are provided to create the divided down clocks used for the I/O Mux/DeMux gearing logic (SCLK inputs of DDR I/O) and they drive the Primary Clock network. There are twelve Edge Clock Dividers on the Nexus device.

5.2. Primary Clock Divider (PCLKDIV)

For CrossLink-NX, Certus-NX, and MachXO5-NX, one programmable Primary Clock Divider is provided to create the divided down clocks. For CertusPro-NX, two programmable Primary Clock Dividers are available.

5.3. Dynamic Clock Select (DCS)

The dynamic clock select provides run-time selectable glitchless or non-glitchless operation between two independent clock sources to the primary clock network. This clock select allows the selection of clock sources without leaving the dedicated clock resources in the device. There is one dynamic clock select block on the CrossLink-NX, Certus-NX, and MachXO5-NX devices, and there are two dynamic clock select blocks on the CertusPro-NX device.

5.4. Dynamic Clock Control (DCC)

Dynamic Clock Control allows dynamic clock enable and disables the MIDMUX Feed Line and the four special common interface block (CIB) clocks from the core. When a Feed Line is disabled, all the logic and clock signals that are fed by this Feed Line do not toggle. Hence, it reduces the overall dynamic power consumption of the device.

5.5. Edge Clock Sync (ECLKSYNC)

The Nexus devices have dynamic edge clock synchronization control (ECLKSYNC) which allows each edge clock to be disabled or enabled glitchlessly from core logic if desired. This allows you to synchronize the edge clock to an event or external signal, if desired. It also allows the design to dynamically disable a clock and its associated logic in the design when it is not needed and thus save power.

5.6. Oscillator (OSC)

An internal programmable rate oscillator is provided. The oscillator can be used for FPGA configuration, Soft Error Detect (SED), and as a user logic clock source that is available after FPGA configuration. There is one OSCA on the Nexus device. The oscillator clock output is routed directly to primary clocking.

The oscillator output is not a high-accuracy clock, having a +/- 7% variation in its output frequency. It is mainly used for circuits that do not require a high degree of clock accuracy. Examples of usage are asynchronous logic blocks such as a timer or reset generator, or other logic that require a constantly running clock.

6. Primary Clocks

6.1. Primary Clock Sources

The primary clock network has multiple inputs, called primary clock sources, which can be routed directly to the primary clock routing to clock the FPGA fabric.

The primary clock sources that can connect to the primary clock routing are:

- Dedicated Clock Input Pins
- PLL Outputs
- PCLKDIV/ECLKDIV Outputs
- Internal FPGA Fabric Entries (with minimum general routing)
- SGMII-CDR, SerDes/PCS clocks
- OSC Clock

All potential primary clock sources are multiplexed prior to going to the primary clock routing by a MIDMUX. There are 58 MIDMUX connections and four FPGA fabric connections, 62 total, routed to a multiplexor in the center of the chip called the centermux. From the centermux, primary clocks are selected and distributed to the FPGA fabric.

The maximum number of unique clock sources is:

18 bottom MIDMUX sources + 16 top MIDMUX sources + 12 left MIDMUX sources + 12 right MIDMUX sources + 4 direct FPGA fabric entry points (from general routing) = 62.

The basic clocking structure is shown in [Figure 3.1](#) and [Figure 3.2](#), elaborated in [Appendix A. Primary Clock Sources and Distribution](#).

6.2. Primary Clock Routing

The primary clock routing network is made up of low skew clock routing resources with connectivity to every synchronous element of the device. Primary clock sources are selected at the MIDMUX, then selected in the centermux and distributed on the primary clock routing to clock the synchronous elements in the FPGA fabric. For CrossLink-NX, Certus-NX, and MachXO5-NX, the primary clock routing network is divided into left and right regions. [Figure 6.1](#) is the simplified view of [Figure 3.1](#). For CertusPro-NX, the primary clock routing network is divided into four regions, up-left, up-right, low-left, and low-right. [Figure 6.2](#) is the simplified view of [Figure 3.2](#).

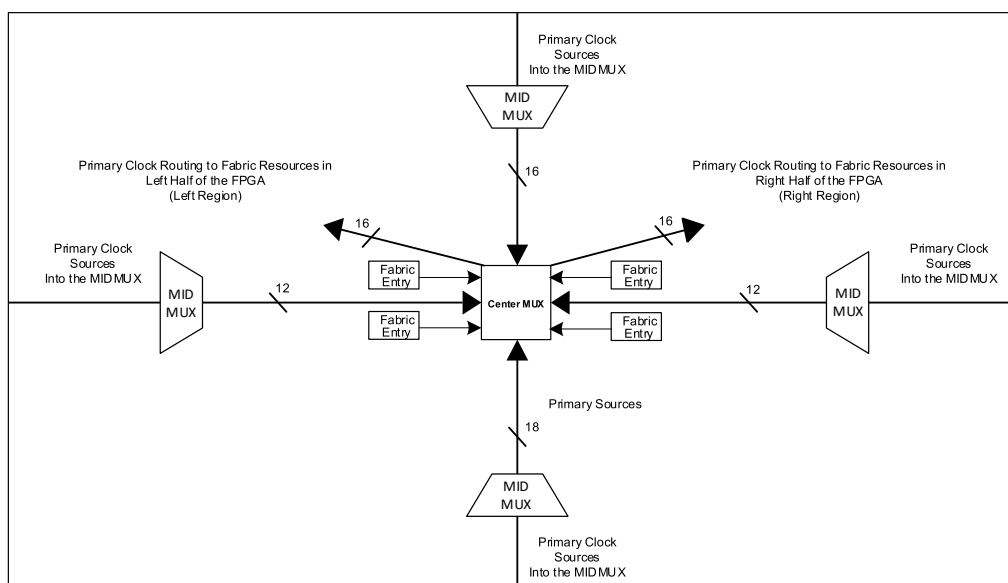


Figure 6.1. Primary Clock Routing Architecture for CrossLink-NX, Certus-NX, and MachXO5-NX

7. Primary Clock Divider (PCLKDIV)

Inside the centermux, one (for CrossLink-NX, Certus-NX, and MachXO5-NX) or two (for CertusPro-NX) Primary Clock Dividers are available. Each Primary Clock Divider provides the following functionalities:

- PCLK Divider supports $\div 2$, $\div 4$, $\div 8$, $\div 16$, $\div 32$, $\div 64$, and $\div 128$. When PCLK divider is bypassed, it is $\div 1$ mode.
- PCLK Divider can be reset by global Reset signals and sleep mode control signals. The global reset can be disabled by a configuration bit.
- PCLK Divider supports user Local Reset through CIB port.
- The reset is Asynchronous assert and synchronous de-assert. The divider output starts at the next cycle after the reset is synchronously released.
- Allow GSR activity to be ignored during device power up by gating this signal with internal DONE.
- When exiting from sleep mode, the retention registers are released from the asynchronous reset control.

7.1. PCLKDIV Component Definition

The PCLKDIV component can be instantiated in the source code of a design as defined in this section. [Figure 7.1](#), [Table 7.1](#), and [Table 7.2](#) define the PCLKDIV component. Verilog and VHDL instantiations are included.

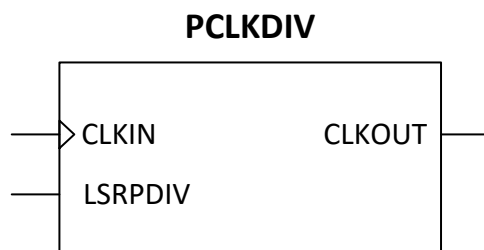


Figure 7.1. PCLKDIV Component Symbol

Table 7.1. PCLKDIV Component Port Definition

| Port Name | I/O | Description |
|-----------|-----|------------------------------------------------------------------------------------------------------------------------------------------------|
| CLKIN | I | Primary Clock Input |
| LSRPDIV | I | Local Reset — Active High, asynchronously forces all outputs low. LSRPDIV = 0 Clock outputs are active LSRPDIV = 1 Clock outputs are OFF |
| CLKOUT | O | Divide by 1, 2, 4, 8, 16, 32, 64, or 128 Output Port |

Table 7.2. PCLKDIV Component Attribute Definition

| Name | Value | Default | Description |
|-------------|-------------------------------------|---------|--------------------------------------|
| DIV_PCLKDIV | X1, X2, X4, X8, X16, X32, X64, X128 | X1 | Primary Clock Divide Ratio Selection |
| GSR | ENABLE DISABLE | ENABLED | GSR ENABLE/DISABLE Selection |

7.2. PCLKDIV Usage in VHDL

Component Instantiation

```
Library lattice;
use lattice.components.all;
```

Component and Attribute Declaration

```
component PCLKDIV
Generic (DIV_PCLKDIV      : string;
        GSR              : string);
Port    (CLKIN   : in STD_LOGIC;
        LSRPDIV  : in STD_LOGIC;
        CLKOUT   : out STD_LOGIC);
end component;
```

PCLKDIV Instantiation

```
attribute DIV_PCLKDIV : string;
attribute DIV_PCLKDIV of I1 : label is "X1";
attribute GSR : string;
attribute GSR of I1 : label is "DISABLED";

I1: PCLKDIV
generic map (DIV_PCLKDIV => "2.0",
            GSR           => "DISABLED")
port map    (CLKIN       => CLKIN,
            LSRPDIV      => LSRPDIV,
            CLKOUT       => CLKOUT);
```

7.3. PCLKDIV Usage in Verilog

Component and Attribute Declaration

```
module PCLKDIV (CLKIN, LSRPDIV, CLKOUT);

parameter DIV_PCLKDIV = "X2";    // "X1", "X2", "X4", "X8", "X16", "X32", "X64", "X128"
parameter GSR = "DISABLED";      // "ENABLED", "DISABLED"

input  CLKIN, LSRPDIV;
output CLKOUT;
endmodule
```

PCLKDIV Instantiation

```
defparam I1.DIV_PCLKDIV = "X2";
defparam I1.GSR = "DISABLED";
PCLKDIV I1 (
    .CLKIN      (CLKIN),
    .LSRPDIV    (LSRPDIV),
    .CLKOUT     (CLKOUT));
```

8. Dynamic Clock Select (DCS)

One (for CrossLink-NX, Certus-NX, and MachXO5-NX) or two (for CertusPro-NX) dynamic clock select (DCS) blocks are located at the center of the PLC array, which can drive to any or all the regions. The DCS_CMUX Structures are shown in Figure 8.1 and Figure 8.2.

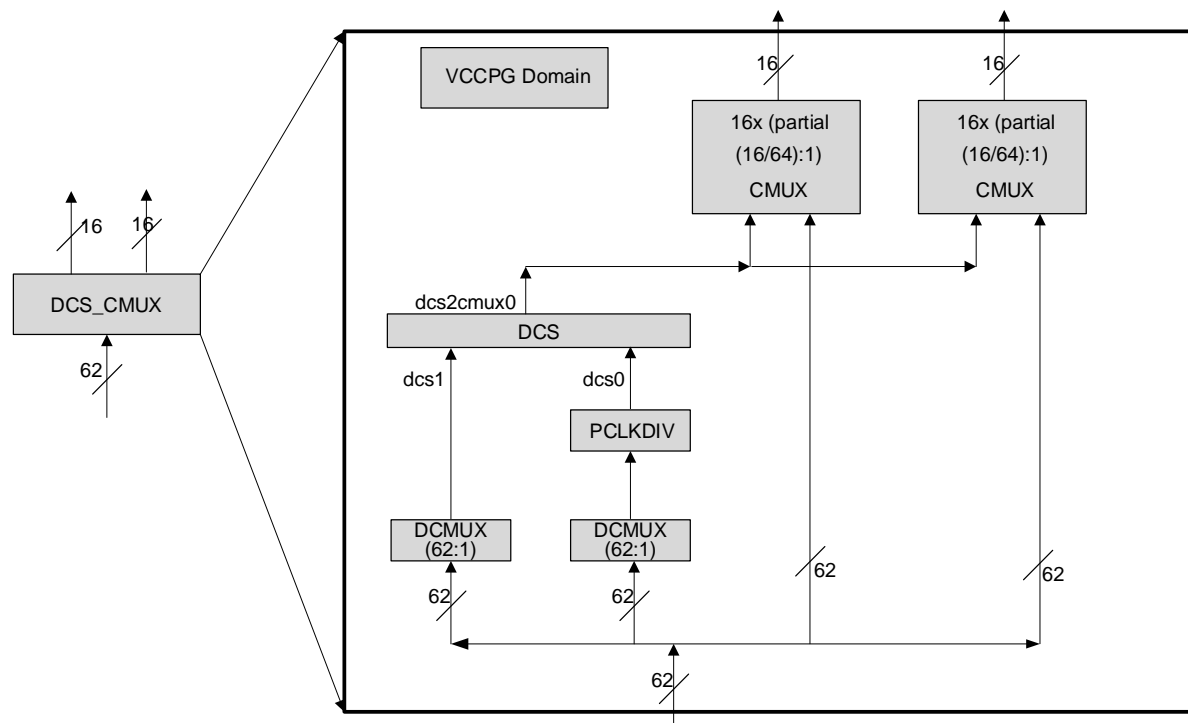


Figure 8.1. DCS_CMUX Structure for CrossLink-NX, Certus-NX, and MachXO5-NX

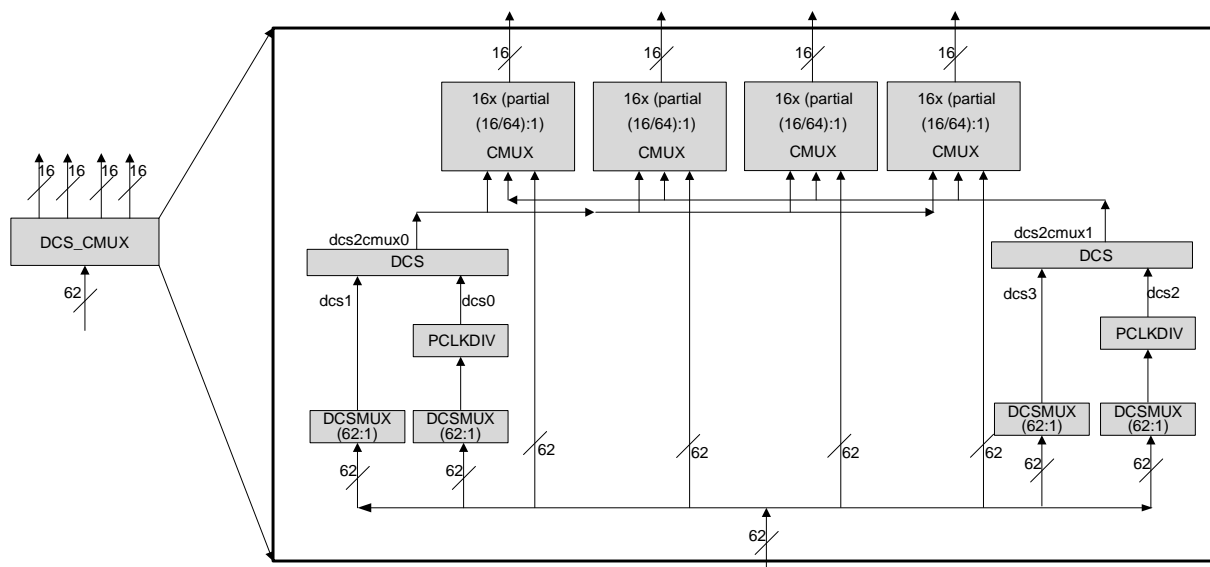


Figure 8.2. DCS_CMUX Structure for CertusPro-NX

The DCS block allows dynamic and glitchless selection between two PCLK clock sources. The DCS block shares the same clock resource as any PCLK CMUX. This way the DCS function can be performed on any two primary clock sources. The inputs to the DCS block come from all the outputs of MIDMUXs and local routing that is located at the center of the PLC array. The output of the DCS is connected to the inputs of Primary Clock Center MUXs. The DCS logic structure is shown in Figure 8.3.

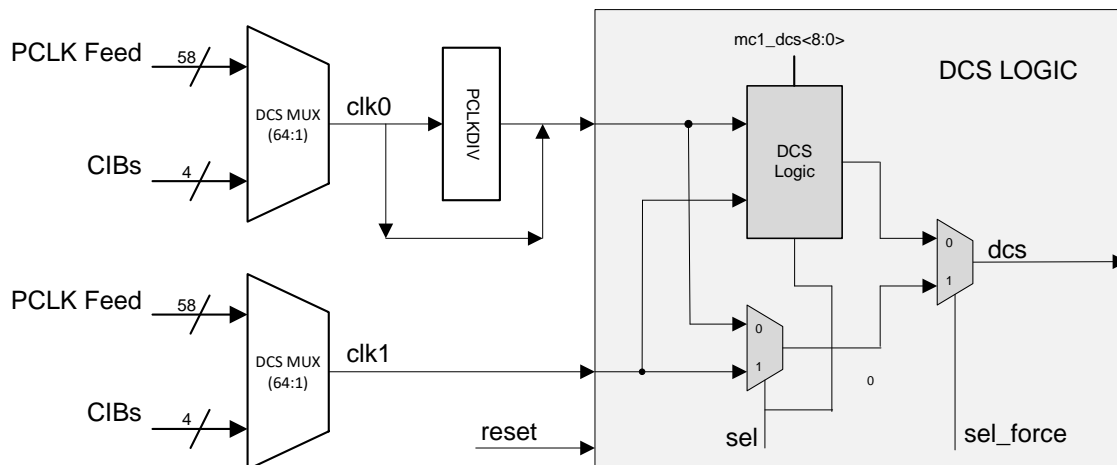


Figure 8.3. DCS Logic Structure

The *DCSMODE* attribute sets the behavior of the DCS output. The DCS attributes are described in Table 8.2.

8.1. DCS Timing Diagrams

The DCS block allows dynamic and glitchless selection between two PCLK clock sources. The DCS block shares the same clock resource as any PCLK CMUX. Therefore, the DCS function can be performed on any two primary clock sources. Figure 8.4, Figure 8.5, and Figure 8.6 show the DCS in glitchless operation in conjunction with the *DCSMODE* attribute. Figure 8.7 shows the non-glitchless bypass operation scenario.

8.1.1. Functionality – posedge SEL switch

The selection switches from current clock to target clock. For posedge configuration, the latch state is low. Below is the sequence of events once SEL toggles:

1. Current clock must see posedge then negedge, then is deactivated.
2. Target clock must see posedge then negedge, then output is successfully switched over.

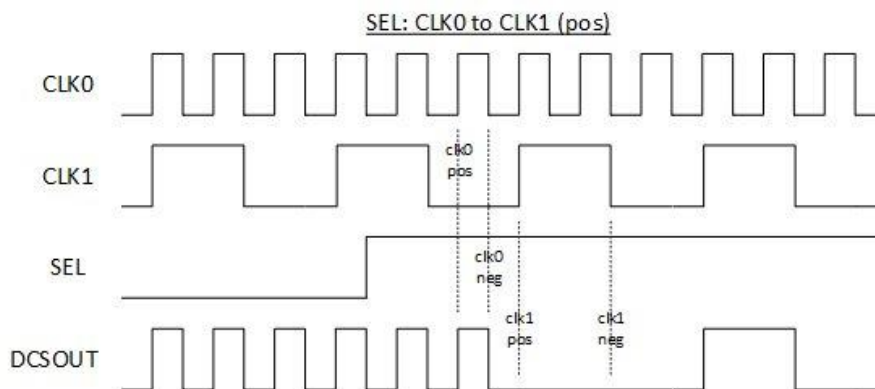


Figure 8.4. Posedge DCS Switch from SEL: 0 => 1

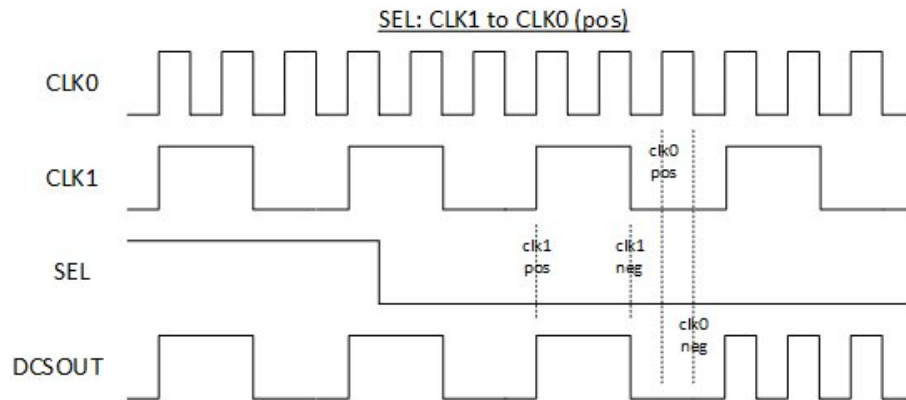


Figure 8.5. Posedge DCS Switch from SEL: 1 => 0

8.1.2. Functionality – negedge SEL switch

The selection switches from current clock to target clock. For negedge configuration, the latch state is high. Below is the sequence of events once SEL toggles:

1. Current clock must see negedge then posedge, then is deactivated.
2. Target clock must see negedge then posedge, then output is successfully switched over.

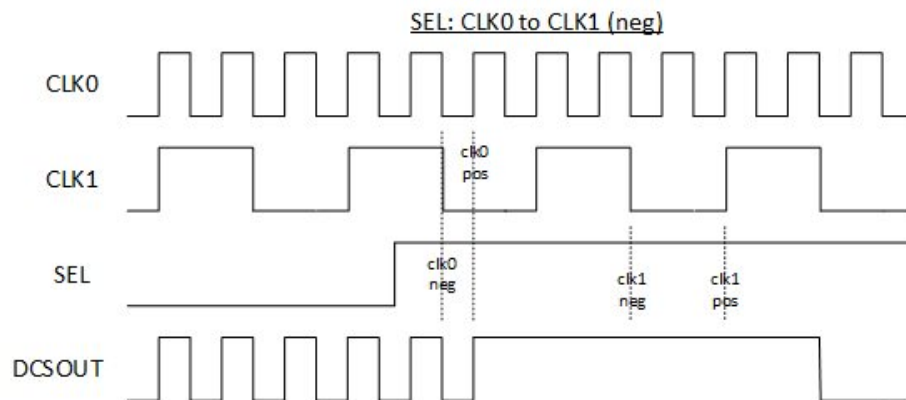


Figure 8.6. Negedge DCS Switch from SEL: 0 => 1

8.1.3. Functionality – bypass

When SELFCE is high, the switch is in bypass mode. The output clock transitions immediately from the current clock to the target clock and may have glitches.

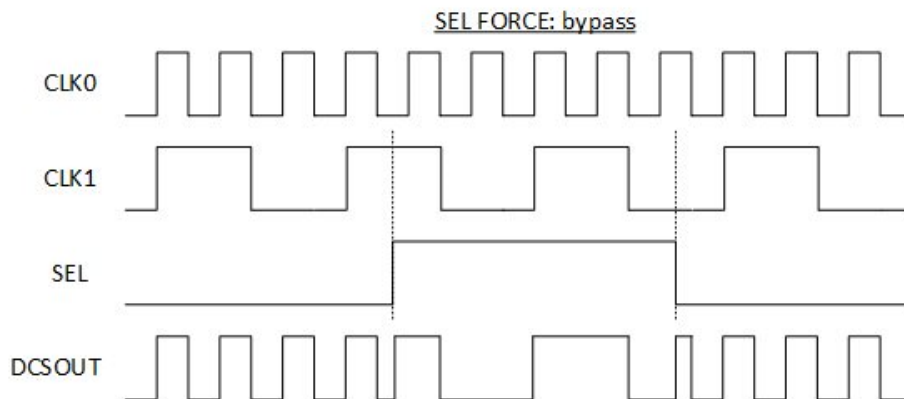


Figure 8.7. SELFCE = 1 DCS Clock Switch

8.2. DCS Component Definition

The DCS component can be instantiated in the source code of a design as defined in this section.

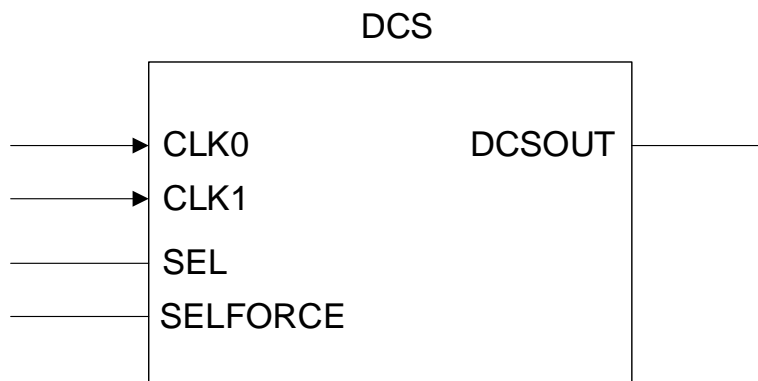


Figure 8.8. DCS Component Symbol

Table 8.1. DCS Component Port Definition

| Port Name | I/O | Description |
|-----------|-----|-------------------------------------------------------|
| CLK0 | I | Clock Input port 0 — Default |
| CLK1 | I | Clock Input port 1 |
| SEL | I | Input Clock Select |
| SELFCE | I | Selects Glitchless (0) or Non-Glitchless (1) behavior |
| DCSOUT | O | Clock Output Port |

Table 8.2 provides the behavior of the DCS output based on the setting of the *DCSMODE* attribute and the SELFCE pin input. The SELFCE pin is dynamic and can toggle during operation. The glitchless switching is only achievable when SELFCE = 0.

Table 8.2. DCS – DCSMODE Attribute

| Attribute Name | Attribute Value | Output | | Description |
|---------------------------|-----------------|---------|---------|-----------------------------------------------|
| | | SEL = 0 | SEL = 1 | |
| DCSMODE (SELFORCE = 0) | VCC | CLK0 | CLK1 | Rising edge triggered. Latched state is high. |
| | GND | CLK0 | CLK1 | Falling edge triggered. Latched state is low. |
| | BUFGCECLK1_0 | 0 | CLK1 | SEL is active high. Disabled output is low |
| | BUFGCECLK1 | 1 | CLK1 | SEL is active high. Disabled output is high. |
| | BUFGCECLK0 | CLK0 | 0 | SEL is active low. Disabled output is low. |
| | BUFGCECLK0_1 | CLK0 | 1 | SEL is active low. Disabled output is high. |
| | BUF0 | CLK0 | CLK0 | Buffer for CLK0 |
| | BUF1 | CLK1 | CLK1 | Buffer for CLK1 |
| SELFORCE= 1 | Non-Glitchless | CLK0 | CLK1 | — |

8.3. DCS Usage in VHDL

Component Instantiation

```
Library lattice;
use lattice.components.all;
```

Component and Attribute Declaration

```
COMPONENT DCS
GENERIC (DCSMODE : string := "VCC");
PORT (CLK0 :IN STD_LOGIC;
      CLK1 :IN STD_LOGIC;
      SEL :IN STD_LOGIC;
      SELFORCE :IN STD_LOGIC;
      DCSOUT :OUT STD_LOGIC);
END COMPONENT;
```

DCS Instantiation

```
attribute DCSMODE : string;
attribute DCSMODE of DCSinst0 : label is "VCC";
I1: DCS
generic map(
  DCSMODE => "VCC")
port map (
  CLK0 => CLK0
  ,CLK1 => CLK1
  ,SEL => SEL
  ,SELFORCE => SELFORCE
  ,DCSOUT => DCSOUT);
```

8.4. DCS Usage in Verilog

Component and Attribute Declaration

```
module DCS (CLK0, CLK1, SEL, SELFORCE, DCSOUT);  
input      CLK0;  
input      CLK1;  
input      SEL;  
input      SELFORCE;  
output     DCSOUT;  
endmodule
```

DCS Instantiation

```
defparam DCSInst0.DCSMODE = "VCC";  
DCS DCSInst0 (  
    .CLK0      (CLK0),  
    .CLK1      (CLK1),  
    .SEL       (SEL),  
    .SELFORCE  (SELFORCE),  
    .DCSOUT    (DCSOUT));
```

9. Dynamic Clock Control (DCC)

The Nexus device has a Dynamic Clock Control feature which allows internal logic to dynamically enable or disable the region primary clock network. This gating function does not create glitches or increase the clock latency to the primary clock network. Also, this dynamic clock control function can be disabled by a configuration memory fuse to always enable the primary clock network.

The DCC controls the clock sources from the Primary CLOCK MIDMUX before they are fed to the Primary Center MUXs that drive the region clock network. When a clock network is disabled, the power consumption of all the associated logic is greatly reduced.

The Nexus device clock architecture allows both DCC and DCS to function at the same time. Care must be taken when the clock source is used as input to the PLL. The DCC should remain enabled, otherwise if the PLL input clock stops toggling, the PLL loses locked and the PLL output clock also stops toggling.

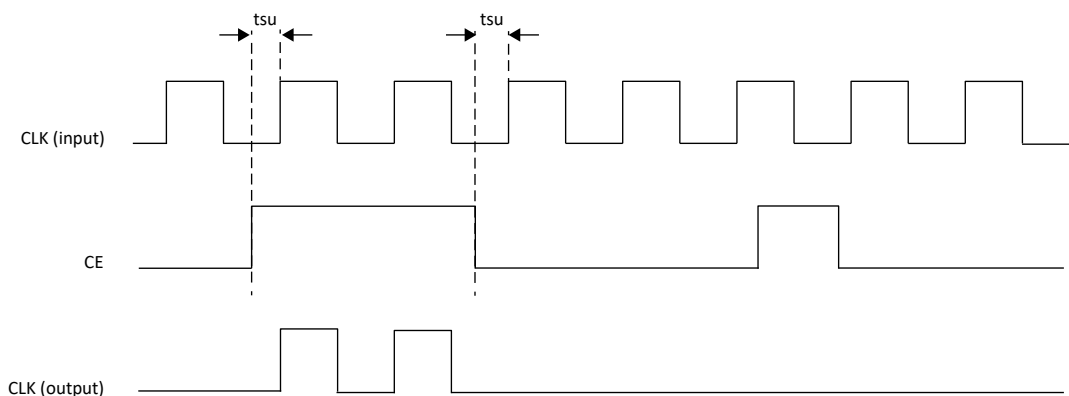


Figure 9.1. Glitchless DCC Functional Waveform

Dynamic Clock Control allows the four clocks from the FPGA fabric feeding to the MIDMUX be dynamically enabled and disabled. When a Feed Line is disabled, all the logic and clock signals that are fed by this Feed Line do not toggle. Hence, it reduces the overall dynamic power.

9.1. Component Definition

The DCC component can be instantiated in the source code of a design as defined in this section. Figure 9.2, Table 9.1, and Table 9.2 show the DCC definitions.

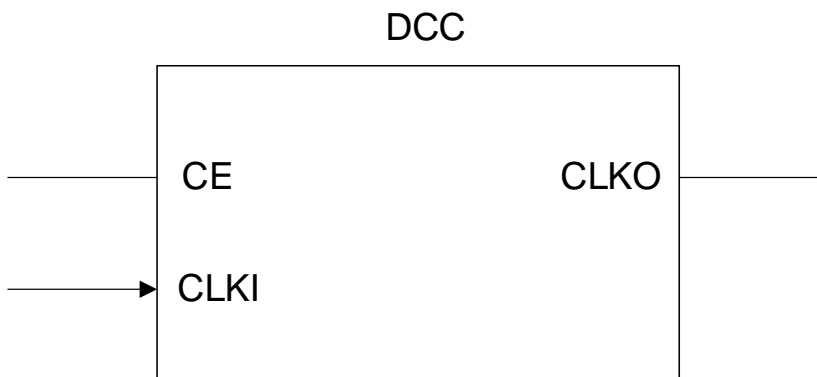


Figure 9.2. DCC Component Symbol

Table 9.1. DCC Component Port Definition

| Port Name | I/O | Description |
|-----------|-----|-------------------------------------------------------------------------------------------------------|
| CLKI | I | Clock Input port. |
| CE | I | Clock Enable port — CE = 0 CLKO is disabled (CLKO = '0') — CE = 1 CLKO is enabled (CLKO = CLKI) |
| CLKO | O | Clock Output Port |

Table 9.2. DCC Component Attribute Definition

| Name | Value | Default | Description |
|-------|--------|---------|----------------------------------------------------------------------------------------------------------------|
| DCCEN | 0 1 | 0 | Enables dynamic control. "0": CLKO = CLKI regardless of the CE input. "1": CLKO depends on the CE input. |

9.2. DCC Usage in VHDL

Component Instantiation

```
library lattice;
use lattice.components.all;
Component and Attribute Declaration
COMPONENT DCC
PORT    (CLKI :IN STD_LOGIC;
         CE    :IN STD_LOGIC;
         CLKO  :OUT STD_LOGIC);
END COMPONENT;
```

DCC Instantiation

```
I1: DCC
port map (
    CLKI  => CLKI,
    CE    => CE,
    CLKO  => CLKO);
DCC Usage in Verilog
Component and Attribute Declaration
module DCC(CLKI,CE,CLKO);
input  CLKI;
input  CE;
output CLKO;
endmodule
```

9.3. DCC Usage in Verilog

DCC Instantiation

```
DCC DCSInst0 (
    .CLKI (CLKI),
    .CE   (CE),
    .CLKO (CLKO));
```

10. Internal Oscillator (OSCA)

The OSCA component performs multiple functions on the Nexus device. It is used for configuration, SED, as well as optionally in user mode. In user mode, the OSCA component has the following features:

- It permits a design to be fully self-clocked, as long as the quality of the OSCA component's silicon-based oscillator is adequate.
- If it is unused, it can be turned off for power savings.
- It has an input to dynamically control standby/normal operation.
- It has a direct connection to primary clock routing through the top MIDMUX. For CertusPro-NX, the right MIDMUX can also be used for the direct connection to primary clock routing.
- It can be configured for operation at a wide range of frequencies through the configuration bits.

10.1. OSCA Component Definition

The OSCA component can be instantiated in the source code of a design as defined in this section. [Figure 10.1](#) and [Table 10.1](#) show the OSCA definitions.

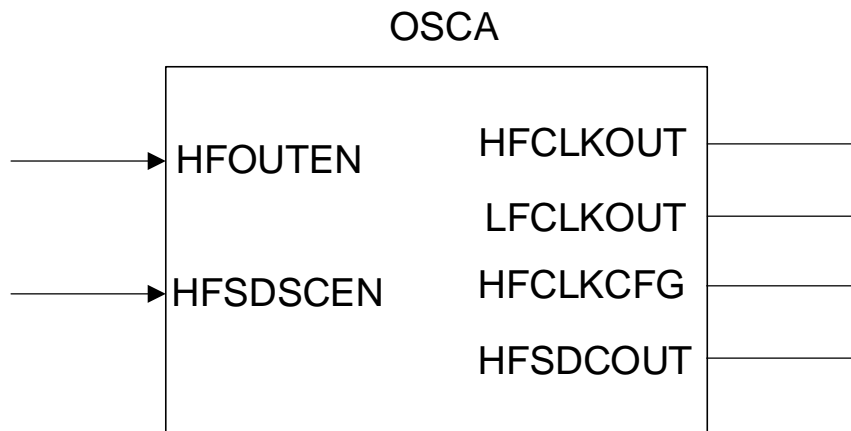


Figure 10.1. OSCA Component Symbol

Table 10.1. OSCA Component Port Definition

| Port Name | I/O | Description |
|-----------|-----|---------------------------------------------------------------------------|
| HFOUTEN | I | High Frequency User Clock Output Enable |
| HFSDSCEN | I | High Frequency User Clock Output Enable |
| HFCLKOUT | O | 450 MHz with Programmable Divider (2~256) to User |
| HFSDCOUT | O | 450 MHz with Programmable Divider (2~256) to User for SED/SEC Application |
| LFCLKOUT | O | Low Frequency Clock Output; 32 kHz |
| HFCLKCFG | O | High Frequency Reference Clock; 450 MHz |

Table 10.2. OSCA Component Attribute Definition

| Name | Value | Default | Description |
|----------------|---------------------------|----------|--------------------------------------------------------------------------|
| HF_CLK_DIV | 00000001 ~ 11111111 | 00000001 | User-assignable HF oscillator output divider configuration (div2~div256) |
| HF_SED_SEC_DIV | 00000001 ~ 11111111 | 00000001 | User-assignable HF oscillator output divider configuration (div2~div256) |
| HF_OSC_EN | DISABLED ENABLED | DISABLED | HF oscillator enable, controlled by the user |
| LF_OUTPUT_EN | DISABLED ENABLED | DISABLED | Low frequency clock output enable |

10.2. OSCA Usage in VHDL

Component Instantiation

```
Library lattice;
use lattice.components.all;
```

Component and Attribute Declaration

```
Component OSCA
generic (
    HF_CLK_DIV : string;
    HF_SED_SEC_DIV : string;
    HF_OSC_EN : string;
    LF_OUTPUT_EN : string
)
port (
    HFOUTEN : in std_logic;
    HFSDSCEN : in std_logic;
    HFCLKOUT : out std_logic;
    LFCLKOUT : out std_logic;
    HFCLKCFG : out std_logic;
    HFSDCOUT : out std_logic
);
```

OSCA Instantiation

```
I1: OSCA
generic map (
    HF_CLK_DIV : "1", --(DIV = 2)
    HF_SED_SEC_DIV : "1", --(DIV = 2)
    HF_OSC_EN : "ENABLED",
    LF_OUTPUT_EN : "ENABLED"
)
port map (
    HFOUTEN => HFOUTEN,
    HFSDSCEN => HFSDSCEN,
    HFCLKOUT => HFCLKOUT,
    LFCLKOUT => LFCLKOUT,
    HFCLKCFG => HFCLKCFG,
    HFSDCOUT => HFSDCOUT
);
```

10.3. OSCA Usage in Verilog

OSCA Instantiation

```
OSCA I1 #(
    .HF_CLK_DIV ("1"), //DIV = 2
    .HF_SED_SEC_DIV ("1"), //DIV = 2
    .HF_OSC_EN ("ENABLED"),
    .LF_OUTPUT_EN ("ENABLED"),
) (
    .HFOUTEN (HFOUTEN ),
    .HFSDSCEN (HFSDSCEN),
    .HFCLKOUT (HFCLKOUT),
    .LFCLKOUT (LFCLKOUT),
    .HFCLKCFG (HFCLKCFG),
    .HFSDCOUT (HFSDCOUT)
);
```

11. Edge Clocks

Each Nexus device bottom I/O bank has four ECLK resources. There are three I/O banks at the bottom of the device. These clocks, which have low injection time and skew, are used to clock I/O registers. Edge Clock resources are designed for high-speed I/O interfaces with high fan-out capability. See [Figure 3.1](#) for ECLK locations and connectivity.

The sources of Edge Clocks are:

- Dedicated Clock (PCLK) pins
- DLLDEL output
- Bottom PLL outputs (CLKOP, CLKOS, CLKOS2, CLKOS3, CLKOS4, and CLKOS5)
- ECLK Bridge
- Internal nodes

The Nexus device has Edge Clock (ECLK) at the bottom of the device. There are four ECLK networks per I/O bank. ECLK Input MUX collects all clock sources available as shown in [Figure 11.1](#). There are three ECLK Input MUXs, one for each I/O bank on the bottom side of the device. Each of these MUX generates total of four ECLK Clock sources for each I/O bank. Each ECLK network from one I/O bank can be bridged to another I/O bank from a wider bus if it is needed.

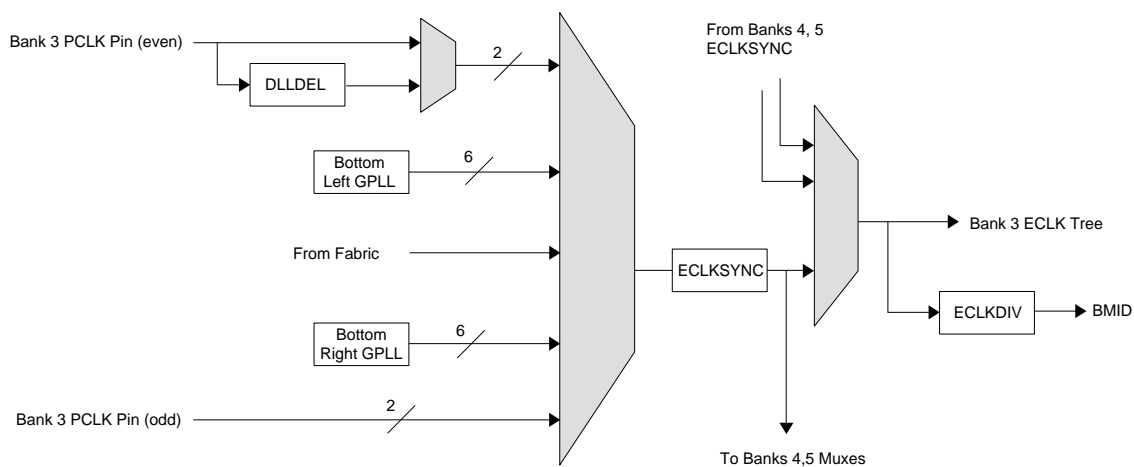


Figure 11.1. Edge Clock Sources Per Bank

11.1. Edge Clock Dividers (ECLKDIV)

There are twelve Edge Clock dividers available in the Nexus device, four for each I/O bank at the bottom of the device. The Clock Divider provides a single divided output with available divide values of 2, 3.5, 4, or 5. The inputs to the Clock Dividers are the Edge Clocks, PLL outputs and Primary Clock Input pins. The outputs of the Clock Divider drive the primary clock network and are mainly used for DDR I/O domain crossing.

11.2. ECLKDIV Component Definition

The ECLKDIV component can be instantiated in the source code of a design as defined in this section. [Figure 11.2](#), [Table 11.1](#), and [Table 11.2](#) define the ECLKDIV component. Verilog and VHDL instantiations are included.



Figure 11.2. ECLKDIV Component Symbol

Table 11.1. ECLKDIV Component Port Definition

| Port Name | I/O | Description |
|-----------|-----|----------------------------------------------------------------------------------------------------------------------------------------------|
| ECLKIN | I | Edge Clock Input |
| DIVRST | I | Reset input — Active High, asynchronously forces all outputs low. DIVRST = 0 Clock outputs are active DIVRST = 1 Clock outputs are OFF |
| SLIP | I | Signal is used for word alignment. When enabled it slips the output one cycle relative to the input clock. |
| DIVOUT | O | Divide by 1, 2, 3.5, 4, or 5 Output Port |

Table 11.2. ECLKDIV Component Attribute Definition

| Name | Value | Default | Description |
|----------|---------------------------------------|---------|-------------------------------------------|
| GSR | ENABLED DISABLED | ENABLED | GSR ENABLE/DISABLE Selection |
| ECLK_DIV | DISABLE "2" "3P5" "4" "5" | DISABLE | ECLK DIVIDE Ratio selection ("3P5" = 3.5) |

The SLIP input is intended for use with high-speed data interfaces such as DDR or 7:1 LVDS Video.

11.3. ECLKDIV Usage in VHDL

Component Instantiation

```
Library lattice;
use lattice.components.all;
```

Component and Attribute Declaration

```
component ECLKDIV
Generic (ECLK_DIV : string;
        GSR       : string);
Port    (DIVRST   : in STD_LOGIC;
        ECLKIN    : in STD_LOGIC;
        SLIP      : in STD_LOGIC;
        DIVOUT     : out STD_LOGIC);
end component;
```

ECLKDIV Instantiation

```
attribute ECLK_DIV : string;
attribute ECLK_DIV of I1 : label is "2.0";
attribute GSR : string;
attribute GSR of I1 : label is "DISABLED";

I1: ECLKDIV
generic map (ECLK_DIV => "2.0",
             GSR      => "DISABLED")
port map    (DIVRST    => DIVRST,
             ECLKIN     => ECLKIN,
             SLIP       => SLIP,
             DIVOUT     => DIVOUT);
```

11.4. ECLKDIV Usage in Verilog

Component and Attribute Declaration

```
module ECLKDIV (DIVRST, ECLKIN, SLIP, DIVOUT);

parameter ECLK_DIV = "2.0";          // "2.0", "3.5"
parameter GSR = "DISABLED"; // "ENABLED", "DISABLED"

input  DIVRST, ECLKIN, SLIP;
output DIVOUT;
endmodule
```

ECLKDIV Instantiation

```
defparam I1.ECLK_DIV = "2.0";
defparam I1.GSR = "DISABLED";
ECLKDIV I1 (
    .DIVRST    (DIVRST),
    .ECLKIN     (ECLKIN),
    .SLIP       (SLIP),
    .DIVOUT     (DIVOUT));
```

12. Edge Clock Synchronization (ECLKSYNC)

Nexus devices have a dynamic Edge Clock synchronization control (ECLKSYNC) which allows each Edge Clock to be disabled or enabled glitchlessly from core logic if desired. This allows you to synchronize the Edge Clock to an event or external signal if desired. It also allows the design to dynamically disable a clock and its associated logic in the design when it is not needed and thus, save power. Applications such as DDR2, DDR3, and 7:1 LVDS for display use this component for clock synchronization.

12.1. ECLKSYNC Component Definition

The ECLKSYNC component can be instantiated in the source code of a design as defined in this section. Asserting the STOP control signal has the ability to stop the Edge Clock to synchronize the signals derived from ECLK and used in high-speed DDR mode applications such as DDR memory, generic DDR, and 7:1 LVDS.

Control signal STOP is synchronized with ECLK when asserted. When control signal STOP is asserted, the clock output is forced to low after the fourth falling edge of the input ECLKI. When the STOP signal is released, the clock output starts to toggle at the fourth rising edge of the input ECLKI clock.

Figure 12.1 and Table 12.1 show the ECLKSYNC component definition.

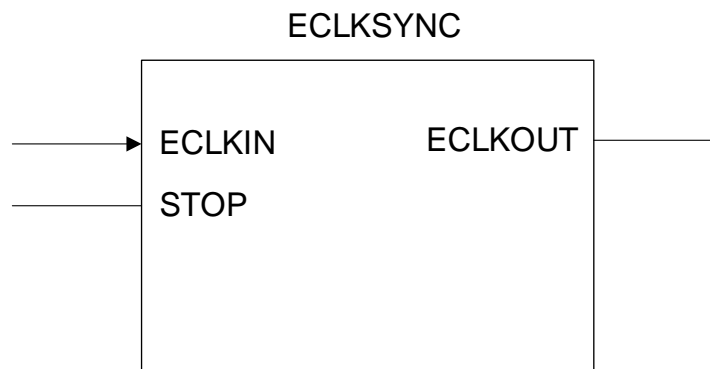


Figure 12.1. ECLKSYNC Component Symbol

Table 12.1. ECLKSYNC Component Port Definition

| Port Name | I/O | Description |
|-----------|-----|---------------------------------------------------------------------------------------------------------------------------------------------|
| ECLKIN | I | Clock Input port. |
| STOP | I | Control signal to stop Edge Clock <ul style="list-style-type: none"> STOP = 0 Clock is Active STOP = 1 Clock is Off |
| ECLKOUT | O | Clock Output Port |

Table 12.2. ECLKSYNC Component Attribute Definition

| Name | Value | Default | Description |
|---------|-------------------|---------|-------------------------------|
| STOP_EN | DISABLE ENABLE | DISABLE | STOP ENABLE/DISABLE Selection |

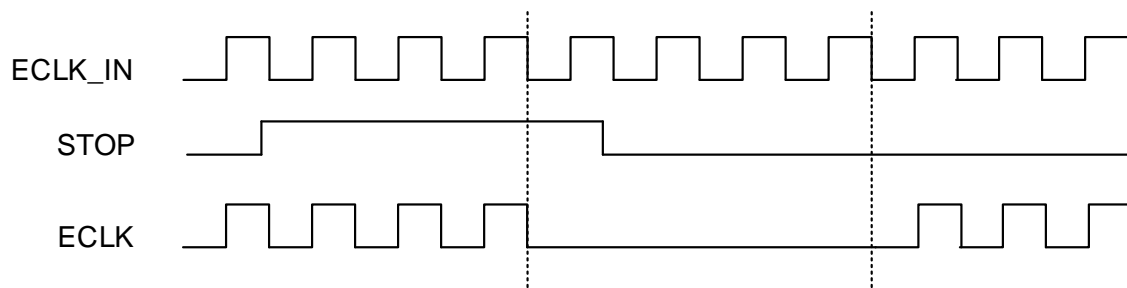


Figure 12.2. ECLKSYNC Functional Waveform

12.2. ECLKSYNC Usage in VHDL

Component Instantiation

```
Library lattice;
use lattice.components.all;
```

Component and Attribute Declaration

```
COMPONENT ECLKSYNC
PORT    (ECLKIN :IN STD_LOGIC;
         STOP   :IN STD_LOGIC;
         ECLKOUT :OUT STD_LOGIC);
END COMPONENT;
```

ECLKSYNC Instantiation

```
I1: ECLKSYNC
port map (
    ECLKIN => ECLKIN,
    STOP   => STOP,
    ECLKOUT => ECLKOUT);
```

12.3. ECLKSYNC Usage in Verilog

Component and Attribute Declaration

```
module ECLKSYNC (ECLKIN, STOP, ECLKOUT);
input  ECLKIN;
input  STOP;
output ECLKOUT;
endmodule
```

ECLKSYNC Instantiation

```
ECLKSYNC ECLKSYNCInst0 (
    .ECLKIN (ECLKIN),
    .STOP   (STOP),
    .ECLKOUT (ECLKOUT));
```

13. General Routing for Clocks

The Nexus device architecture supports the ability to use general routing for a clock. This capability is intended to be used for small areas of the design to allow additional flexibility in linking dedicated clocking resources and building very small clock trees. General routing cannot be used for Edge Clocks for applications that use the DDR registers in the I/O components of the FPGA.

Software limits the distance of a general routing based (gated) clock to one PLC in distance to a primary clock entry point. If the software cannot place the clock gating logic close enough to a primary clock entry point, the error below occurs:

- ERROR-par – Unable to reach a primary clock entry point for general route clock <net> in the minimum required distance of one PLC.

There are multiple entry points to the Primary clock routing throughout the Nexus device fabric. In this case, it is recommended to add a preference for this gated clock to use primary routing.

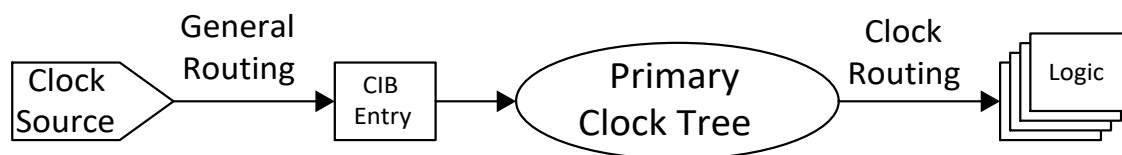


Figure 13.1. Gated Clock to the Primary Clock Routing

For a very small clock domain, you can limit the distance of a general routing based (gated) clock to one PLC in distance to the logic it clocks. You must group this logic (UGROUP) with a *BBOX = 1, 1* (see Lattice Radiant Help > Constraints Reference Guide > Preferences > UGROUP) as well as specify a *PROHIBIT PRIMARY* on the generated clock. If the software cannot place the logic tree within the BBOX, an error occurs.

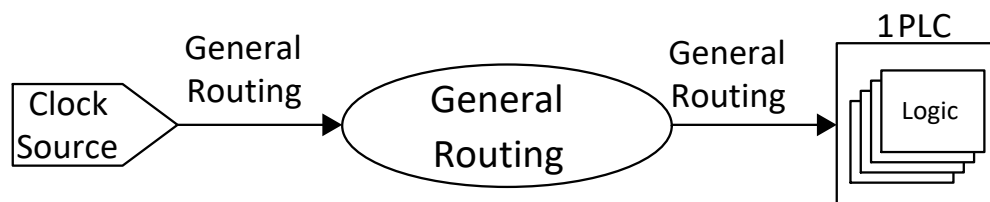


Figure 13.2. Gated Clock to Small Logic Domain

14. sysCLOCK PLL

14.1. sysCLOCK PLL Overview

The sysCLOCK™ PLLs can be used in a variety of clock management applications such as clock injection delay removal, clock phase adjustment, clock timing adjustment, and frequency synthesis (multiplication and division of a clock). The PLL supports Fractional-N synthesis. The Nexus IP Catalog PLL user interface shows important timing parameters such as the VCO rate and the PLL loop bandwidth.

The PLL Input sources are:

- Dedicated PLL Input Pins. See [Appendix A](#) for more details.
- Primary Clock Routing
- Edge Clock Routing
- FPGA Fabric

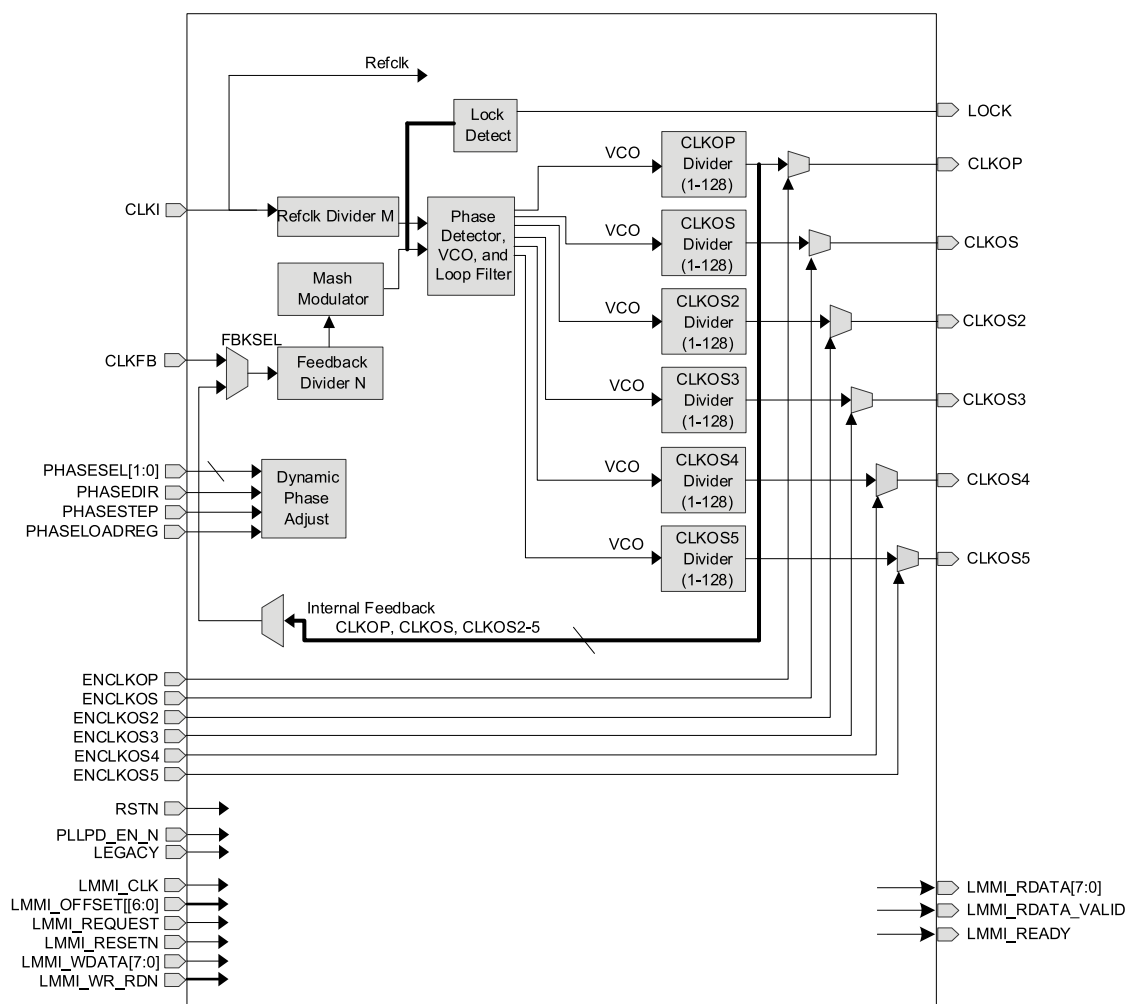


Figure 14.1. Nexus PLL Block Diagram

There are three PLLs for LIFCL-40 and LFD2NX-40 at three corners as Upper Left, Lower Left and Lower Right, two PLLs for LIFCL-17 and LFD2NX-17 at two corners as Lower Left and Lower Right corners. There are three PLLs for LFCPNX-50 at three corners as Upper Left, Lower Left and Lower Right, four PLLs for LFCPNX-100 at four corners as Upper Left, Upper Right, Lower Left and Lower Right. There are two PLLs for LFMXO5 at two corners as Upper Left and Lower Right. There is one PLL for LIFCL-33 at Lower Left corner. Each PLL has six outputs. All six PLL outputs can feed the Primary Clock and Edge Clock networks.

14.2. PLL Features

14.2.1. Dedicated PLL Inputs

Every PLL has a dedicated low skew input (PLLCK) that routes directly to its reference clock input. These are the recommended inputs for a PLL. It is possible to route a PLL input from the Primary clock routing, but it incurs more clock input injection delays, which are not natively compensated for using feedback, compared to a dedicated PLL input. In each corner of one Nexus device, there is one PLL at most. Each PLL on the Nexus device has one pair of dedicated PLL input pins.

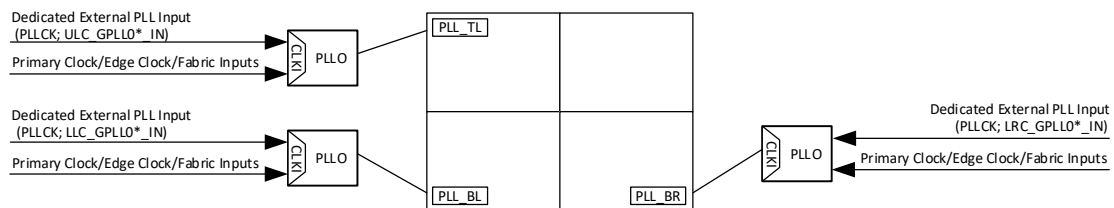


Figure 14.2. PLL Input Pins for LIFCL-40 and LFD2NX-40

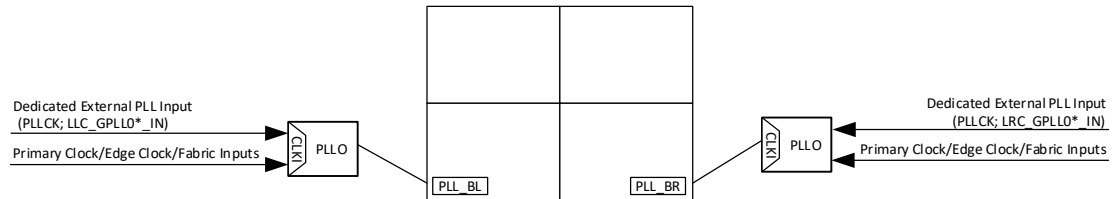


Figure 14.3. PLL Input Pins for LIFCL-17 and LFD2NX-17

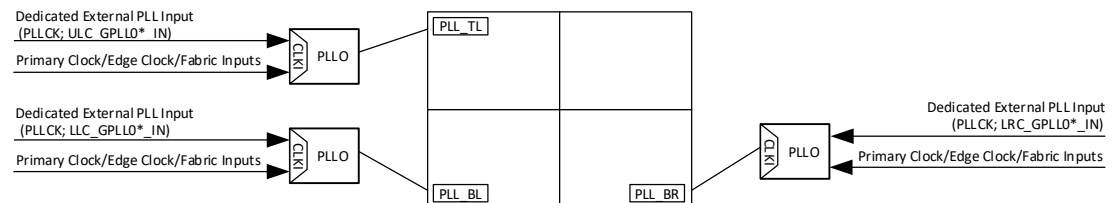


Figure 14.4. PLL Input Pins for LFCPNX-50

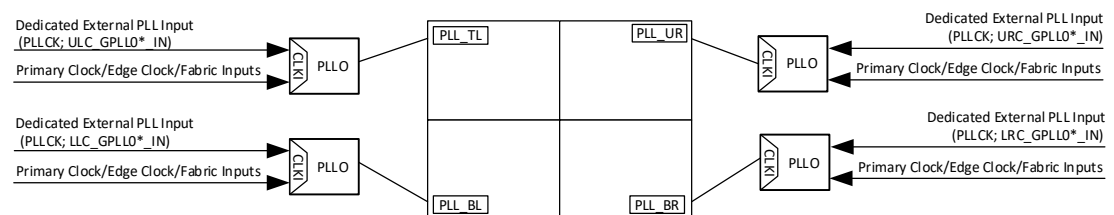


Figure 14.5. PLL Input Pins for LFCPNX-100

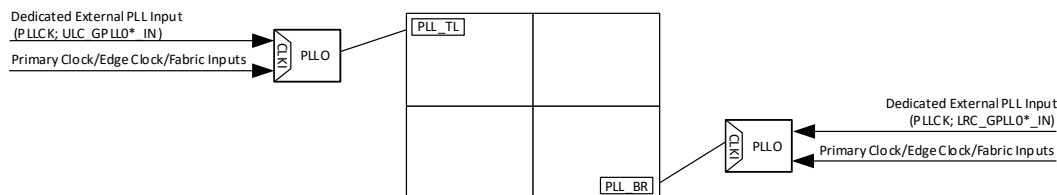


Figure 14.6. PLL Input Pins for LFMXO5

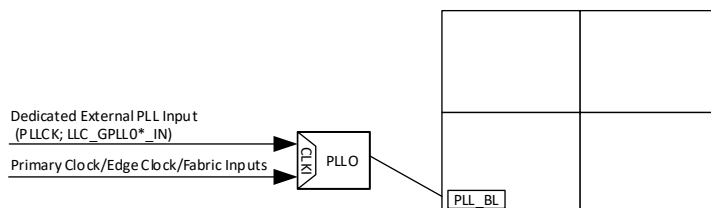


Figure 14.7. PLL Input Pins for LIFCL-33

14.2.2. Clock Injection Delay Removal

The clock injection delay removal feature of the PLL removes the delay associated with the PLL and clock tree. This feature is typically used to reduce the clock path delay which benefits system synchronous input and output timing. This feature is performed by aligning the PLL input clock with a feedback clock from the clock tree. Optional delay may also be added to the feedback path to further reduce the clock injection time.

14.2.3. Clock Phase Adjustment

The clock phase adjustment feature of the PLL provides the ability to set a specific phase offset between the outputs of the PLL. New to the Nexus device, phase adjustments can be calculated in much finer increments since the frequency is used to calculate the available phase increments. This feature is detailed further in the Dynamic Phase Adjustment section.

14.2.4. Frequency Synthesis

The PLL can be used to multiply up or divide down an input clock.

14.2.5. Legacy Mode (Standby)

In addition to the major features, the PLL has a Legacy Mode to reduce power. The Legacy Mode was called PLL standby mode. But due to the new proposed scheme for Nexus PLLs, it is given a different name to differentiate with the new STDBY mode. The Legacy Mode allows the PLL to be placed into a standby state to save power when not needed in the design. Standby mode is very similar to holding the PLL in reset since the VCO is turned off and needs to regain lock when exiting standby. In both cases, reset and standby mode, the PLL retains its programming.

You MUST hold the PLL in standby for a minimum of 1 ms in order to be sure the PLL analog circuits are fully reset and analog startup is stable.

14.3. sysCLOCK PLL Component Definition

The PLL component can be instantiated in the source code of a design as defined in this section. [Figure 14.8](#) and [Table 14.1](#) show the definitions.

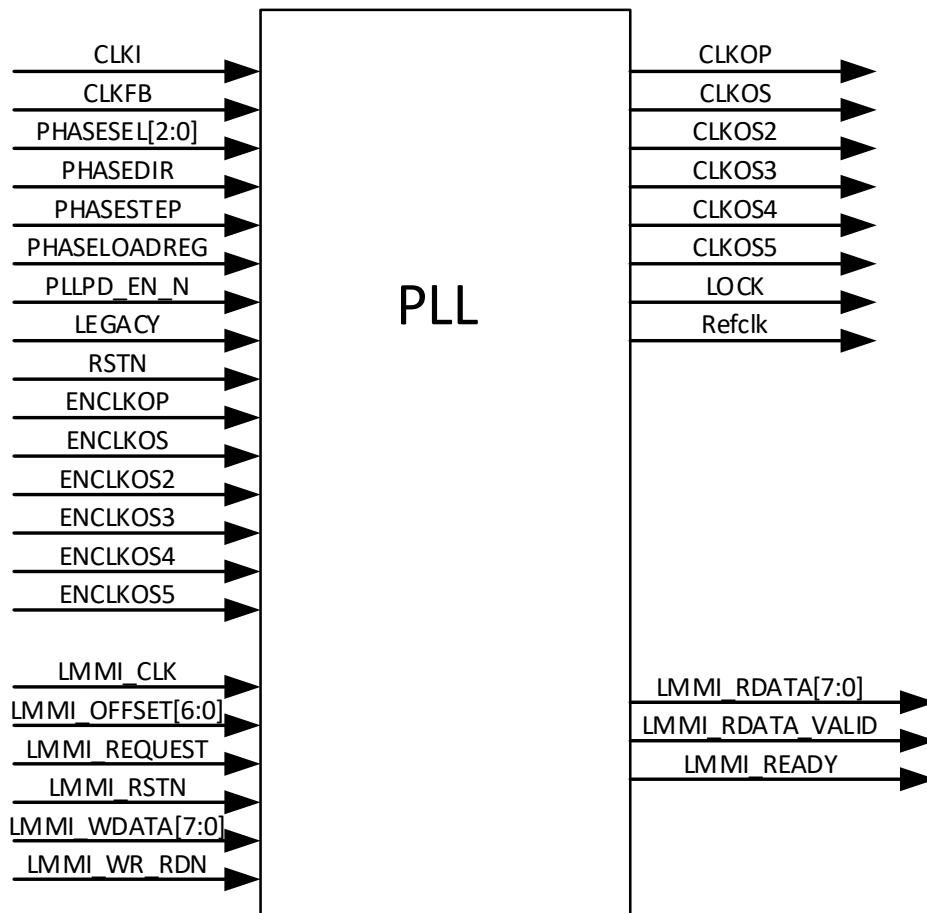


Figure 14.8. PLL Component Instance

Table 14.1. PLL Component Port Definition

| Signal | I/O | Description |
|------------------|-----|-------------------------------------------------------------------------------------------|
| CLKI | I | Input Clock to PLL. |
| CLKFB | I | Feedback Clock. |
| PHASESEL[2:0] | I | Select the output affected by Dynamic Phase adjustment. |
| PHASEDIR | I | Dynamic Phase adjustment direction. |
| PHASESTEP | I | Dynamic Phase adjustment step. |
| PHASELOADREG | I | Load dynamic phase adjustment values into PLL. |
| PLLPD_EN_N | I | Standby signal to power down the PLL. |
| LEGACY | I | Power mode setting to enable legacy mode |
| RST | I | Resets the entire PLL. |
| ENCLKOP | I | Enable PLL output CLKOP. |
| ENCLKOS | I | Enable PLL output CLKOS. |
| ENCLKOS2 | I | Enable PLL output CLKOS2. |
| ENCLKOS3 | I | Enable PLL output CLKOS3. |
| ENCLKOS4 | I | Enable PLL output CLKOS4. |
| ENCLKOS5 | I | Enable PLL output CLKOS5. |
| CLKOP | O | PLL main output clock. |
| CLKOS | O | PLL output clock. |
| CLKOS2 | O | PLL output clock2. |
| CLKOS3 | O | PLL output clock3. |
| CLKOS4 | O | PLL output clock4. |
| CLKOS5 | O | PLL output clock5. |
| LOCK | O | Indicates PLL is now locked to CLKI, Asynchronous signal. Active high indicates PLL lock. |
| Refclk | O | Output of Reference clock. |
| LMMI_CLK | I | CIB LMMI interface clock |
| LMMI_OFFSET[6:0] | I | CIB LMMI interface address offset (LSB of address bus) |
| LMMI_REQUEST | I | CIB LMMI interface request signal |
| LMMI_RESETN | I | CIB LMMI interface reset, active low |
| LMMI_WDATA[7:0] | I | CIB LMMI interface write data |
| LMMI_WR_RDN | I | CIB LMMI interface Write/Read control; 1=write, 0=read. |
| LMMI_RDATA[7:0] | O | CIB LMMI interface read data |
| LMMI_RDATA_VALID | O | CIB LMMI interface read data valid signal |
| LMMI_READY | O | CIB LMMI interface ready signal |

14.4. Functional Description

14.4.1. Refclk (CLKI) Divider

The CLKI divider is used to control the input clock frequency into the phase detector. The valid PLL input frequency range is specified in the device data sheet.

14.4.2. Feedback Loop (CLKFB) Divider

The CLKFB divider is used to divide the feedback signal, effectively multiplying the output clock. The VCO block increases the output frequency until the divided feedback frequency equals the input frequency. The output of the feedback divider must be within the phase detector frequency range specified in the device data sheet. This port is only available to user interface when *user clock* option is selected for feedback clock. Otherwise, this port is connected by the tool to the appropriate signal you selected in the software.

14.4.3. Output Clock Dividers (CLKOP, CLKOS, CLKOS2, CLKOS3, CLKOS4, CLKOS5)

The output Clock Dividers allow the VCO frequency to be scaled up to the maximum range to minimize jitter. Each of the output dividers is independent of the other dividers and each uses the VCO as the source by default. Each of the output dividers can be set to a value of 1 to 128.

14.4.4. Phase Adjustment (Static Mode)

The CLKOP, CLKOS, CLKOS2, CLKOS3, CLKOS4, and CLKOS5 outputs can be phase adjusted relative to the enabled unshifted output clock. New to the Nexus devices, phase adjustments are now calculated values in the software tools based on VCO clock frequency. This provides a finer phase shift depending on the required frequency. The clock output selected as the feedback cannot use the static phase adjustment feature since it causes the PLL to unlock.

14.4.5. Phase Adjustment (Dynamic Mode)

The phase adjustments can also be controlled in a dynamic mode using the PHASESEL, PHASEDIR, PHASESTEP, and PHASELOADREG ports. See the [Dynamic Phase Adjustment](#) section for usage details. The clock output selected as the feedback cannot use the dynamic phase adjustment feature since it causes the PLL to unlock.

Similar restrictions apply to other clocks.

14.5. PLL Inputs and Outputs

14.5.1. CLKI Input

The CLKI signal is the reference clock for the PLL. It must conform to the specifications in the data sheet for the PLL to operate correctly. The CLKI signal can come from a dedicated PLL input pin or from internal routing. The dedicated dual-purpose I/O pin provides a low skew input path and is the recommended source for the PLL. The reference clock can be divided by the input (M) divider to create one input to the phase detector of the PLL.

14.5.2. CLKFB Input

The CLKFB signal is the feedback signal to the PLL. The feedback signal is used by the Phase Detector inside the PLL to determine if the output clock needs adjustment to maintain the correct frequency and phase. The CLKFB signal can come from a primary clock net (feedback mode = CLKOP[P/S/S2/S3/S4/S5]) to remove the primary clock routing injection delay, from a dedicated external dual-purpose I/O pin (feedback mode = UserClock) to account for board level clock alignment, or from an internal PLL connection (feedback mode = INT_O[P/S/S2/S3/S4/S5]) for simple feedback. The feedback clock signal is divided by the feedback (N) divider to create an input to the phase detector of the PLL. A bypassed PLL output cannot be used as the feedback signal.

14.5.3. RST Input

At power-up, an internal power-up reset signal from the configuration block resets the PLL. Additionally, an active high, asynchronous, user-controlled reset port can be optionally added to the PLL. The RST signal can be driven by an internally generated reset function or by an I/O pin. This RST signal resets the PLL core (VCO, phase detector, and charge pump) and the output dividers which causes the outputs to be logic 0. In bypass mode, the output does not reset.

After the RST signal is deasserted, the PLL starts the lock-in process and takes tLOCK time, about 16 ms, to complete. [Figure 14.9](#) shows the timing diagram of the RST input. The RST signal is active high. The RST signal is optional. Trst = 1 ms reset pulse width, Trstrec = 1 ns time after a reset before the divider output starts counting again.

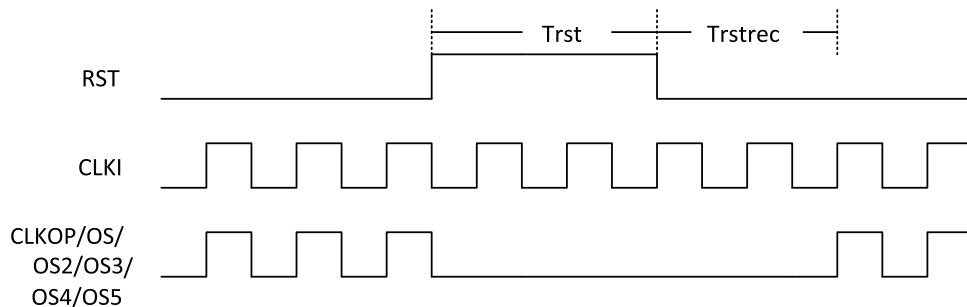


Figure 14.9. RST Input Timing Diagram

14.5.4. Dynamic Clock Enables

Each PLL output has a user input signal to dynamically enable/disable its output clock glitchlessly. When the clock enable signal is set to logic 0, the corresponding output clock is held to logic 0.

Table 14.2. PLL Clock Output Enable Signal List

| Clock Enable Signal Name | Corresponding PLL Output | IP Catalog Option Name |
|--------------------------|--------------------------|------------------------|
| ENCLKOP | CLKOP | "Clock Enable OP" |
| ENCLKOS | CLKOS | "Clock Enable OS" |
| ENCLKOS2 | CLKOS2 | "Clock Enable OS2" |
| ENCLKOS3 | CLKOS3 | "Clock Enable OS3" |
| ENCLKOS4 | CLKOS4 | "Clock Enable OS4" |
| ENCLKOS5 | CLKOS5 | "Clock Enable OS5" |

The Dynamic Clock Enable function allows you to save power by stopping the corresponding output clock when not in use. The clock enable signals are optional and are only available if you select the corresponding option in IP Catalog Wizard. If a clock enable signal is not requested, its corresponding output is active at all times when the PLL is instantiated unless the PLL is placed into standby mode. You cannot access a PLL output clock enable signal in IP Catalog Wizard when the PLL output is used for external feedback to avoid shutting off the feedback clock.

14.5.5. PLLPD_EN_N Input

The PLLPD_EN_N signal is used to put the PLL into a low power standby mode when it is not required. The PLLPD_EN_N signal is optional and is only available if you select the *Enable Powerdown Mode* in the IP Catalog wizard. The PLLPD_EN_N signal is active low. When asserted, the PLL outputs are pulled to 0 and the PLL is reset. You need to stay in the Power Down mode for at least 1 ms to make sure the PLL analog circuits are fully reset and to have a stable analog startup.

14.5.6. Dynamic Phase Shift Inputs

The Nexus PLL has five ports to allow for dynamic phase adjustment from FPGA logic. The Dynamic Phase Adjustment section elaborates on how you should drive these ports.

14.5.7. PHASESEL Input

The PHASESEL[2:0] inputs are used to specify which PLL output port is affected by the dynamic phase adjustment ports. The settings available are shown in the [Dynamic Phase Adjustment](#) section. The PHASESEL signal must be stable for 5 ns before the PHASESTEP or PHASELOADREG signals are pulsed. The PHASESEL signal is optional and is available if you select the *Enable Dynamic Phase Ports* option in IP Catalog Wizard.

Table 14.3. PHASESEL Signal Settings Definition

| PHASESEL[2:0] | PLL Output Shifted |
|---------------|--------------------|
| 000 | CLKOP |
| 001 | CLKOS |
| 010 | CLKOS2 |
| 011 | CLKOS3 |
| 100 | CLKOS4 |
| 101 | CLKOS5 |

14.5.8. PHASEDIR Input

The PHASEDIR input is used to specify which direction the dynamic phase shift occurs, advanced (leading) or delayed (lagging). When PHASEDIR = 0, then the phase shift is delayed. When PHASEDIR = 1, then the phase shift is advanced. The PHASEDIR signal must be stable for 5 ns before the PHASESTEP or PHASELOADREG signals are pulsed. The PHASEDIR signal is optional and is available if you select the *Enable Dynamic Phase Ports* option in IP Catalog Wizard.

Table 14.4. PHASEDIR Signal Settings Definition

| PHASEDIR | Direction |
|----------|--------------------|
| 0 | Delayed (lagging) |
| 1 | Advanced (leading) |

14.5.9. PHASESTEP Input

The PHASESTEP signal is used to initiate a VCO dynamic phase shift for the clock output port and in the direction specified by the PHASESEL and PHASEDIR inputs. This phase adjustment is done by changing the phase of the VCO in 45° increments. The VCO phase changes on the negative edge of the PHASESTEP input after four VCO cycles. This is an active low signal and the minimum pulse width (both high and low) of PHASESTEP pulse is four VCO cycles. The PHASESTEP signal is optional and is available if you select the *Enable Dynamic Phase Ports* option in IP Catalog Wizard. The PHASESEL and PHASEDIR are required to have a setup time of 5 ns prior to PHASESTEP falling edge.

14.5.10. PHASELOADREG Input

The PHASELOADREG signal is used to initiate a post-divider dynamic phase shift, relative to the unshifted output, for the clock output port specified by PHASESEL and in the direction specified by the PHASESEL and PHASEDIR inputs. A phase shift is started on the falling edge of the PHASELOADREG signal and there is a minimum pulse width of 10 ns from assertion to desertion. The PHASESEL and PHASEDIR are required to have a setup time of 5 ns prior to PHASELOADREG falling edge. The PHASELOADREG signal is optional and is available if you select the *Enable Dynamic Phase Ports* option in IP Catalog Wizard.

14.5.11. PLL Clock Outputs

The PLL has six outputs, listed in Table 14.5. All six outputs can be routed to the Primary clock routing of the FPGA. All six outputs can be phase shifted statically or dynamically if external feedback on the clock is not used. They can also statically or dynamically adjust their output duty cycle. The outputs can come from their output divider or the reference clock input (PLL bypass). In bypass mode, the output divider can be bypassed or used to divide the reference clock.

Table 14.5. PLL Clock Outputs and ECLK Connectivity

| Clock Output Name | Edge Clock Connectivity | Selectable Output |
|-------------------|-------------------------|-------------------------------|
| CLKOP | ECLK Connection | Always Enabled |
| CLKOS | ECLK Connection | Selectable through IP Catalog |
| CLKOS2 | No ECLK Connection | Selectable through IP Catalog |
| CLKOS3 | No ECLK Connection | Selectable through IP Catalog |
| CLKOS4 | No ECLK Connection | Selectable through IP Catalog |
| CLKOS5 | No ECLK Connection | Selectable through IP Catalog |

14.5.12. LOCK Output

The LOCK output provides information about the status of the PLL. After the device is powered up and the input clock is stable, the PLL achieves lock within 16 ms. Once lock is achieved, the PLL LOCK signal is asserted. The LOCK signal can be set in IP Catalog Wizard in either the default *unsticky* frequency lock mode by checking the *Provide PLL Lock Signal* or sticky lock mode by selecting *PLL Lock is Sticky*. In sticky lock mode, once the LOCK signal is asserted (logic 1), it stays asserted until a PLL reset is asserted. In the default lock mode of *unsticky* frequency lock, if during operation the input clock or feedback signals to the PLL become invalid, the PLL loses lock and the LOCK output de-asserts (logic 0). It is recommended to assert PLL RST to re-synchronize the PLL to the reference clock when the PLL loses lock. The LOCK signal is available to the FPGA routing to implement the generation of the RST signal if requested by the designer. The LOCK signal is optional and is available if you select the Provide PLL Lock Signal option in IP Catalog Wizard.

14.6. Dynamic Phase Adjustment

Dynamic phase adjustment of the PLL output clocks can be done without reconfiguring the FPGA by using the dedicated dynamic phase-shift ports of the PLL.

All six output clocks, CLKOP, CLKOS, CLKOS2, CLKOS3, CLKOS4, and CLKOS5 have the dynamic phase adjustment feature but only one output clock can be adjusted at a time. Table 14.3 shows the output clock selection settings available for the PHASESEL[2:0] signal. The PHASESEL signal must be stable for 5 ns before the PHASESTEP or PHASELOADREG signals are pulsed.

The selected output clock phase is either advanced or delayed depending upon the value of the PHASEDIR port. Table 14.4 shows the PHASEDIR settings available. The PHASEDIR signal must be stable for 5 ns before the PHASESTEP or PHASELOADREG signals are pulsed.

14.6.1. VCO Phase Shift

Once the PHASESEL and PHASEDIR have been set, a VCO phase adjustment is made by toggling the PHASESTEP signal from the current setting. Each pulse of the PHASESTEP signal generates a phase step based on this equation:

$$\text{VCO Shifted Phase per step} = [1 / (8 \times (\text{DIVO}_{\langle n \rangle_ACTUAL_STR} + 1))] \times 360^\circ$$

Where $\langle n \rangle$ is the clock output specified by PHASESEL (CLKOP/OS/OS2/OS3/OS4/OS5). Values for $\text{DIVO}_{\langle n \rangle_ACTUAL_STR}$ are located in the HDL source file generated by IP Catalog Wizard.

The PHASESTEP signal is latched in on the falling edge and is subject to a minimum wait of four VCO cycles prior to pulsing the signal again. One step size is the smallest phase shift that can be generated by the PLL in one pulse. The dynamic phase adjustment results in a glitch free adjustment when delaying the output clock, but glitches may result when advancing the output clock.

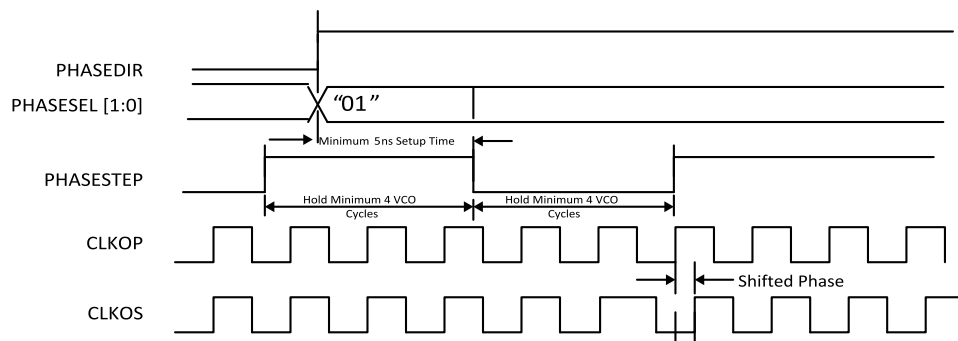


Figure 14.10. PLL Phase Shifting Using the PHASESTEP Signal

For Example:

PHASESEL[2:0]=3'b001 to select CLKOS for phase shift

PHASEDIR =1'b0 for selecting delayed (lagging) phase

Assume the output is divided by 2, DIVOS_ACTUAL_STR = 1

The above signals need to be stable for 5 ns before the falling edge of PHASESTEP and the minimum pulse width of PHASESTEP should be four VCO clock cycles. It should also stay low for four VCO Clock Cycles.

For each toggling of PHASESTEP, you are getting $[1/(8 \times 2)] \times 360 = 22.5$ degree phase shift (delayed).

14.6.2. Divider Phase Shift

Once the PHASESEL and PHASEDIR have been set a post-divider phase adjustment is made by toggling the PHASELOADREG signal. For a desired post-divider phase, it could be calculated by this equation:

$$\text{Post-Divider Phase} = [(\text{DEL}\langle n \rangle - \text{DIVO}\langle n \rangle_ACTUAL_STR) / (\text{DIVO}\langle n \rangle_ACTUAL_STR + 1)] \times 360^\circ$$

Where $\langle n \rangle$ is the clock output specified by PHASESEL (CLKOP/OS/OS2/OS3/OS4/OS5). Values for DEL $\langle n \rangle$ and DIVO $\langle n \rangle_ACTUAL_STR$ are located in the HDL source file generated by IP Catalog Wizard. Please note that if these values are both 1, no shift is made.

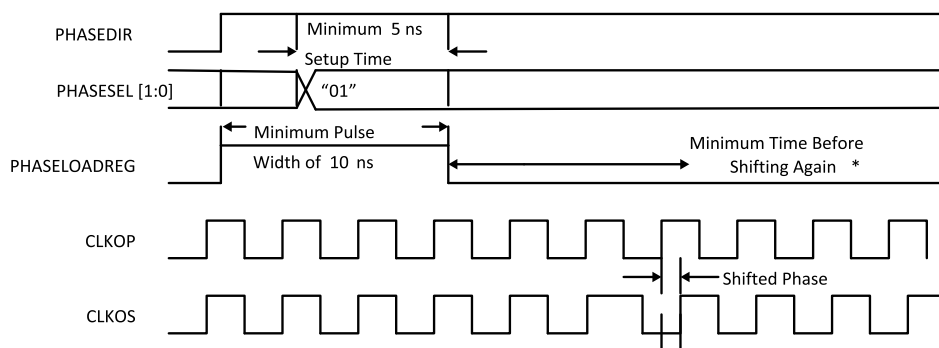


Figure 14.11. Divider Phase Shift Timing Diagram

*Note: Minimum Time Before Shifting Again is:

$$[2.5 \times (\text{DIVO}\langle n \rangle_ACTUAL_STR + 1) + (\text{DEL}\langle n \rangle + 1)] \times (\text{Period of Divider Clock}).$$

14.6.3. Total Phase Shift

For the total phase shift calculation for each PLL output section:

Total Phase Shift = VCO Phase <n> + Post Divider Phase <n>
Post divider Phase <n> = $360 \times (\text{DEL}<n> - \text{DIV}<n>) / (\text{DIV}<n> + 1)$
VCO Phase <n> = $45 \times ((\text{PHI}<n> + m - n) / (\text{DIV}<n> + 1))$; where PHI is the initial Phase shift; m is the number of toggles when dir=0; n is the number of toggles when dir=1;

14.7. Fractional-N Synthesis Operation

The Nexus PLL supports high resolution (12-bit) fractional-N synthesis through Radiant IP Catalog. Fractional-N frequency synthesis allows the user to generate an output clock which is a non-integer multiple of the input frequency. The Fractional- N synthesis option is enabled in the IP Catalog user interface by checking the *Enable fractional-N Divider* box under the *General* tab with the *Configuration Mode* set in either *Frequency* mode or *Divider* mode. When enabled, Fractional-N synthesis is applied to all active PLL outputs.

In *Frequency* configuration mode, user needs to set the CLKI Frequency (10 MHz – 800 MHz) and the desired output Frequency Desired Value (6.25 MHz – 800 MHz) with reasonable Tolerance (0.5% - 10%). Then the Feedback Divider Actual Value (integer), Feedback Divider Actual Value (Fractional) and the clock output *Divider Actual Value* is automatically set. The *Frequency Actual Value* and *ERROR* (PPM) are automatically calculated.

In *Divider* configuration mode, user needs to set the CLKI Frequency, CLKI Divider Desired Value, CLKFB FBK Divider Desired Value (Integer), CLKFB FBK Divider Desired Value (Fractional) and the clock output *Divider Desired Value*. Then, the output clock *Frequency Actual Value* is calculated automatically in the user interface.

The output frequency is given by the equation:

$$F_{out} = \frac{F_{CLKI}}{M \times O} \times (N + \frac{F}{4096})$$

Where:

F_{out} is the output *Frequency Actual Value*.

F_{CLKI} is the CLKI input frequency.

M is the CLKI *Divider Desired Value*.

N is the CLKFB FBK *Divider Desired Value (Integer)*.

F is the CLKFB FBK *Divider Desired Value (Fractional)*.

O is the output *Divider Actual Value*.

The Fractional-N synthesis works by using a delta-sigma technique to approximate the fractional value that was entered by the user. Therefore, using the Fractional-N synthesis option results in higher jitter of the PLL VCO and output clocks compared to using an integer value for the feedback divider. It is recommended that Fractional-N synthesis only be used if the N/M divider ratio is 4 or larger to prevent impacting the PLL jitter performance excessively.

14.8. Spread Spectrum Clock Generation

The Nexus PLL supports Spread spectrum clock generation through Radiant IP Catalog. The spread spectrum function is integrated with the Fractional-N controls and supports *Centered Spread* or *Down Spread*, triangle wave, 0.25% per step from 1.00% to 2.00% with modulation frequency range from 24.42 kHz to 200 kHz. The Spread Spectrum Clock Generation is enabled in the IP Catalog user interface by checking the *Enable Spread Spectrum Clock Generation* box under the *General* tab. In the *Spread Spectrum* section, select *Spread Spectrum Profile* for *Centered Spread* or *Down Spread*, set the *Triangle Modulation Depth* (1.00% - 2.00% with 0.25% step size) and the *Desired Modulation Frequency* (24.42 kHz – 200 kHz). When enabled, spread spectrum characteristics is applied to all active PLL outputs.

Figure 14.12 and Figure 14.13 show the spread spectrum profiles.

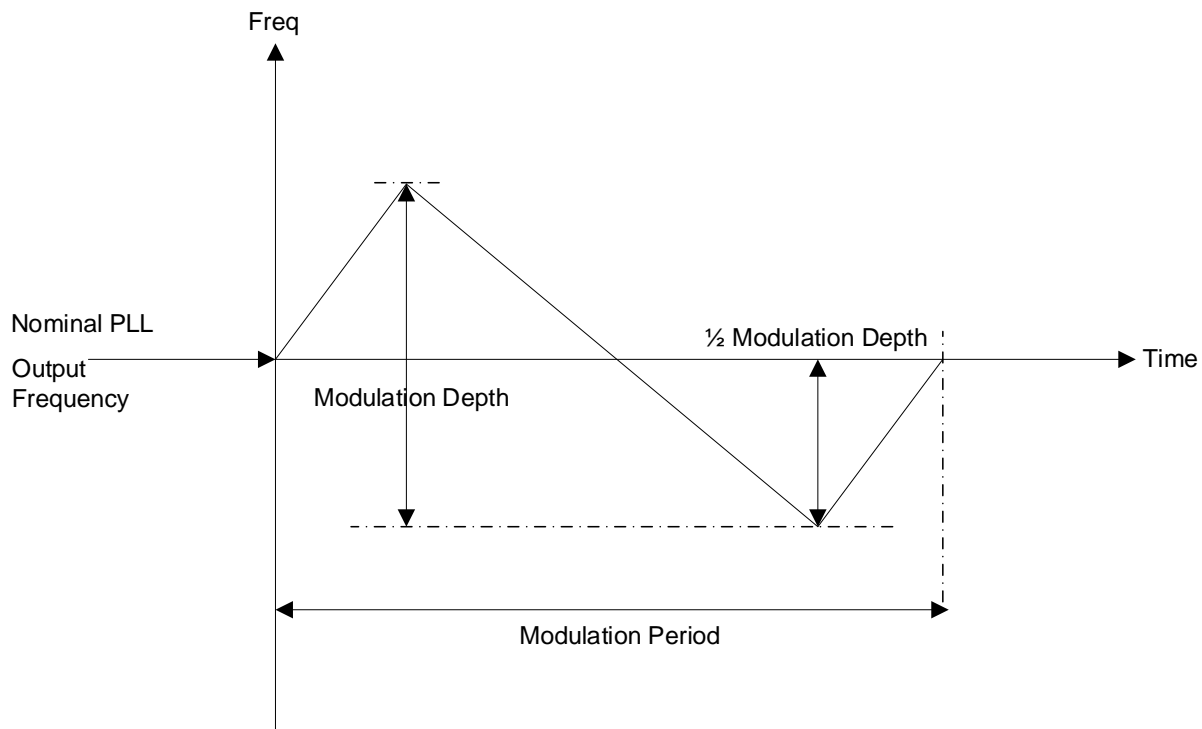


Figure 14.12. Center Spread Profile

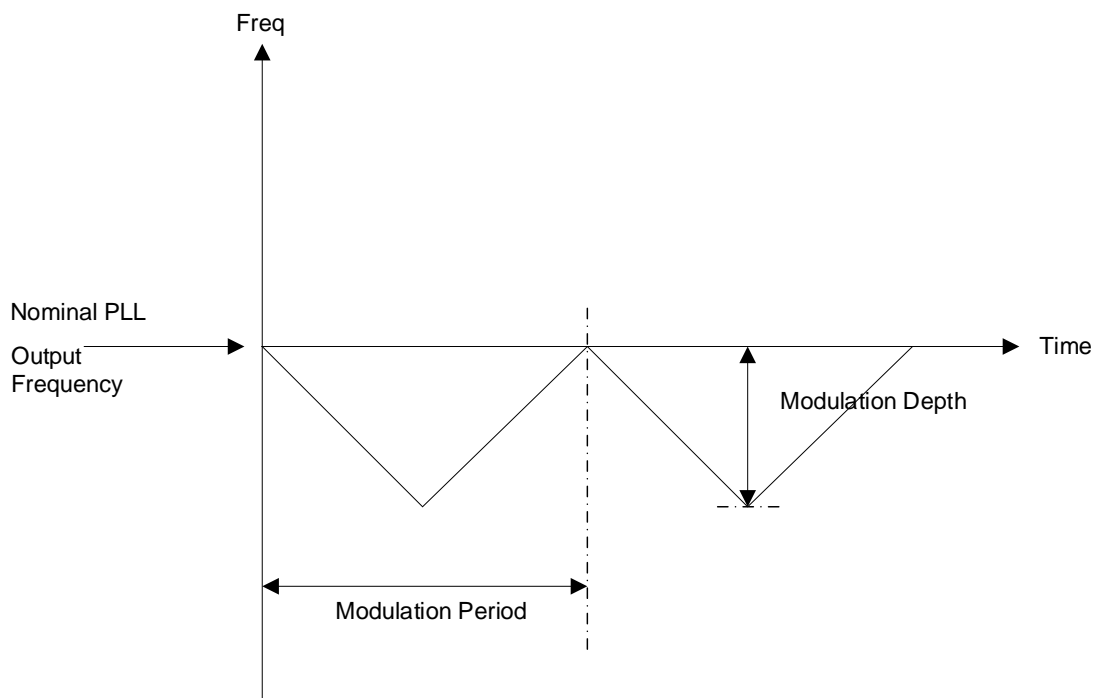


Figure 14.13. Down Spread Profile

14.9. Low Power Features

The Nexus PLL contains several features that allows you to reduce the power usage of a design including Standby mode support and Dynamic clock enable.

14.9.1. Dynamic Clock Enable

The Dynamic Clock Enable feature allows you to glitchlessly enable and disable selected output clocks during periods when not used in the design. A disabled output clock is logic 0. Re-enabled clocks start on the falling edge of the associated clocks. To support this feature, each output clock has an independent Output Enable signal that can be selected. The Output Enable signals are ENCLKOP, ENCLKOS, ENCLKOS2, ENCLKOS3, ENCLKOS4, and ENCLKOS5. Each clock enable port has an option in the IP Catalog user interface to bring the signal to the top level ports of the PLL. If external feedback is used on a port or if the clock output is not enabled, its dynamic clock enable port is unavailable.

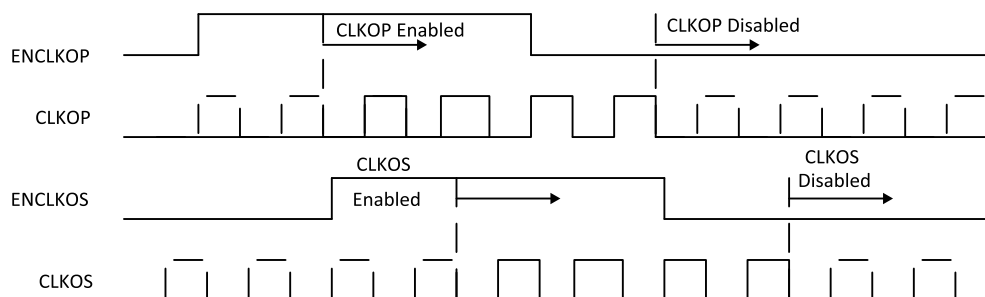


Figure 14.14. Dynamic Clock Enable for PLL Outputs

14.10. PLL Usage in IP Catalog

IP Catalog is used to create and configure a PLL. PLL can be found in the IP Catalog under Module - Architecture Modules. The graphical user interface is used to select parameters for the PLL. The result is an HDL block to be used in the simulation and synthesis flow.

The main window when the PLL is selected is shown in Figure 14.15. When opening IP Catalog inside a Lattice Radiant project, the only entry required is the file name as the other entries are set to the project settings. After entering the module name of choice, click Next to open the PLL configuration window as shown in Figure 14.15.

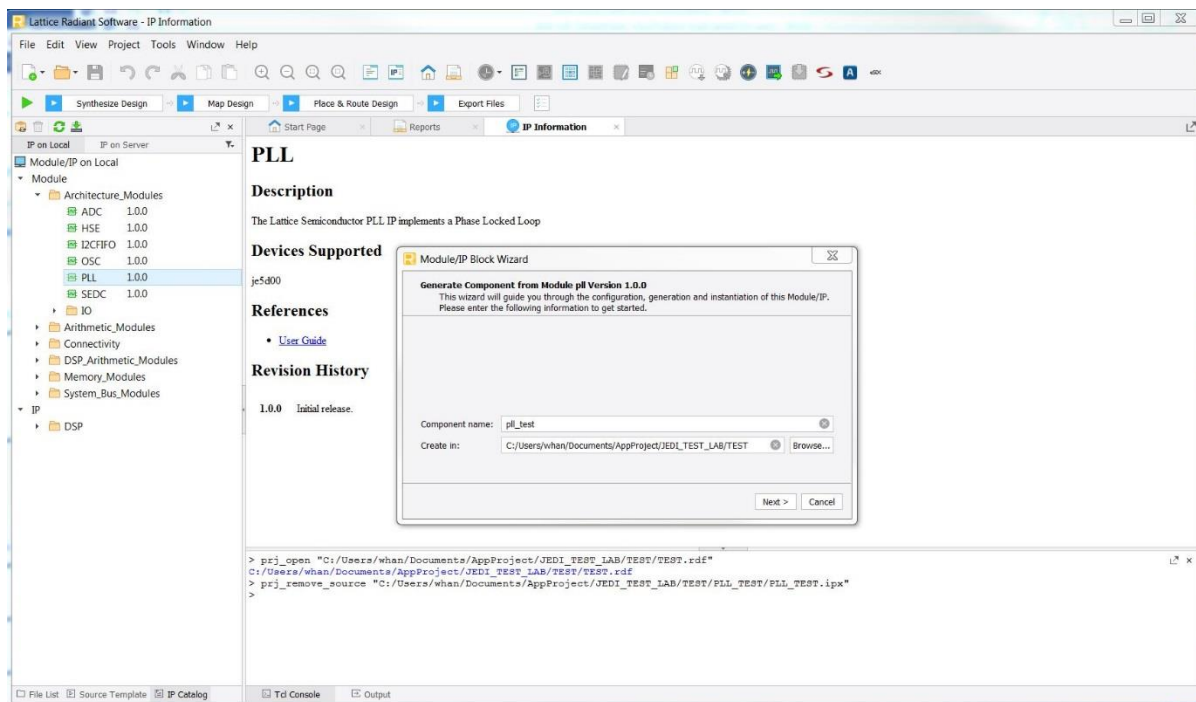


Figure 14.15. IP Catalog Main Window for PLL Module

14.10.1. Configuration Tab

The configuration window lists all user accessible attributes with default values set. Upon completion, click Generate to generate the source.

14.10.2. PLL Frequency and Phase Configuration

In the General Tab, enter the input and output clock frequencies and the software calculates the divider settings. If an entered value is out of range, it is displayed in red and an error message is displayed. You can also select a tolerance value from the *Tolerance %* drop-down box.

If required, enter the desired phase shift and click the Calculate button. The software calculates the closest achievable phase shift and displays it in the *Actual Phase* text box. If an entered value is out of range, it is displayed in red and an error message is displayed.

General Tab

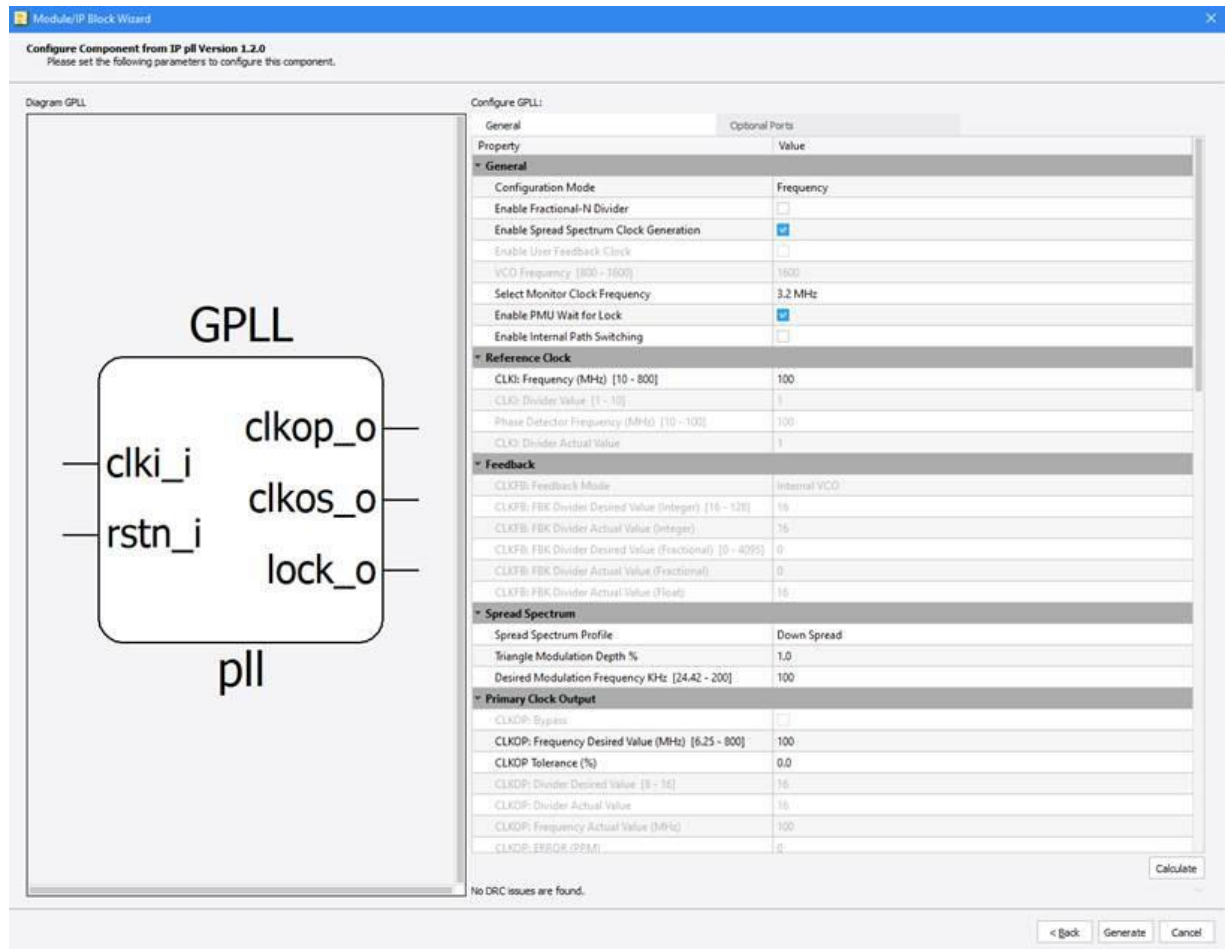


Figure 14.16. Nexus PLL Frequency Configuration in General Tab

Table 14.6. Tab 1, General Settings, IP Catalog User Interface

| User Parameters | Range | Default | Description |
|-----------------------------------------------|---------------------------------------------------------------------------------------------------------------|-------------|------------------------------------------------------------------------------------------------------------------------------------------------------------------------|
| General | | | |
| Configuration Mode | Frequency, Divider | Frequency | Select the configuration mode. Frequency – set the desired input and output frequencies. Divider – set the desired input frequency and desired divider settings. |
| Enable Fractional-N Divider | Checked, Unchecked | Unchecked | Enable/Disable the Fractional Feedback Clock Divider. |
| Enable Spread Spectrum Clock Generation | Checked, Unchecked | Unchecked | Enable/Disable the Spread Spectrum Clock Generation. |
| Enable User Feedback Clock | Checked, Unchecked | Unchecked | When enabled, feedback clock is from user input. |
| VCO Frequency | Calculated | N/A | Display only. |
| Select Monitor Clock Frequency | 3.2 MHz, 1.0 MHz | 3.2 MHz | Select the frequency for reference clock monitoring logic. |
| Enable Internal Path Switching | Checked, Unchecked | Unchecked | Enable/Disable the internal path switching during POR/Sleep/Standby modes. |
| Reference Clock | | | |
| CLKI: Frequency (MHz) | 10–800 | 100 | Set the Reference Clock frequency. (applicable for Frequency mode only) |
| CLKI: Divider Value | 1–128 | 1 | Set the Reference Clock divider. (applicable for Divider mode only) |
| Phase Detector Frequency (MHz) | Calculated | N/A | Display only. |
| CLKI: Divider Actual Value | Calculated | N/A | Display only. |
| Feedback | | | |
| CLKFB: Feedback Mode | CLKOP, CLKOS, CLKOS2, CLKOS3, CLKOS4, CLKOS5, INTCLKOP, INTCLKOS, INT_CLKOS2, INTCLKOS3, INTCLKOS4, INTCLKOS5 | CLKOP | Select the feedback clock from the enabled PLL clock outputs (internal or external). |
| CLKFB: FBK Divider Desired Value (Integer) | 1–128 | 1 | Set the Feedback Clock divider. (applicable for Divider mode only) |
| CLKFB: FBK Divider Actual Value (Integer) | Calculated | N/A | Display only. |
| CLKFB: FBK Divider Desired Value (Fractional) | 0 to 4095 | 0 | Set the Feedback Clock fractional divider. (applicable if Fractional-N Divider is enabled) |
| CLKFB: FBK Divider Actual Value (Fractional) | Calculated | N/A | Display only. |
| CLKFB: FBK Divider Actual Value (Float) | Calculated | N/A | Display only (Integer + Fractional). |
| Spread Spectrum | | | |
| Spread Spectrum Profile | Down Spread, Center Spread | Down Spread | Select the Spread Spectrum Profile. (applicable if Spread Spectrum Clock Generation is enabled) |

| User Parameters | Range | Default | Description |
|-------------------------------------------|------------------------------------|-----------|---------------------------------------------------------------------------------------------------|
| Triangle Modulation Depth % | 0.25, 0.5, 0.75,...,2.0 | 1.0 | Select the modulation depth. (applicable if Spread Spectrum Clock Generation is enabled) |
| Desired Modulation Frequency kHz | 24.42 kHz–200 kHz | 100 | Set the desired modulation frequency. (applicable if Spread Spectrum Clock Generation is enabled) |
| Clock Output | | | |
| CLKO*: Enable | Checked, Unchecked | Unchecked | Enable/Disable PLL Clock Output |
| CLKO*: Bypass | Checked, Unchecked | Unchecked | Bypass the actual divider output and output the reference clock instead. |
| CLKO*: Frequency Desired Value (MHz) | 6.25–800 MHz | 100 | Set the Output Clock frequency. (applicable for Frequency mode only) |
| CLKO*: Tolerance (%) | 0, 0.1, 0.2, 0.5, 1, 2, 5, 10 | 0.0 | Set the acceptable tolerance for actual vs desired output frequency. |
| CLKO*: Divider Desired Value | 1–128 | 8 | Set the Output Clock frequency. (applicable for Frequency mode only) |
| CLKO*: Divider Actual Value | Calculated | N/A | Display Only. |
| CLKO*: Frequency Actual Value (MHz) | Calculated | N/A | Display Only. |
| CLKO*: Static Phase Shift (Degrees) | 0, 45, 90, 135, 180, 225, 270, 315 | 0 | Set the desired clock output phase. |
| CLKO*: ERROR (PPM) | Calculated | 0 | Display Only. Difference between desired and actual frequencies. |
| CLKO*: Enable Trim for CLKO* | Checked, Unchecked | Unchecked | Enable/Disable Trim for clock output. |
| CLKO*: Duty Trim Options Mode | Rising, Falling | Falling | Select Trim mode. |
| CLKO*: Duty Trim Options Delay Multiplier | 0, 1, 2, 4 | 0 | Select Trim Delay Multiplier. |

Optional Ports Tab

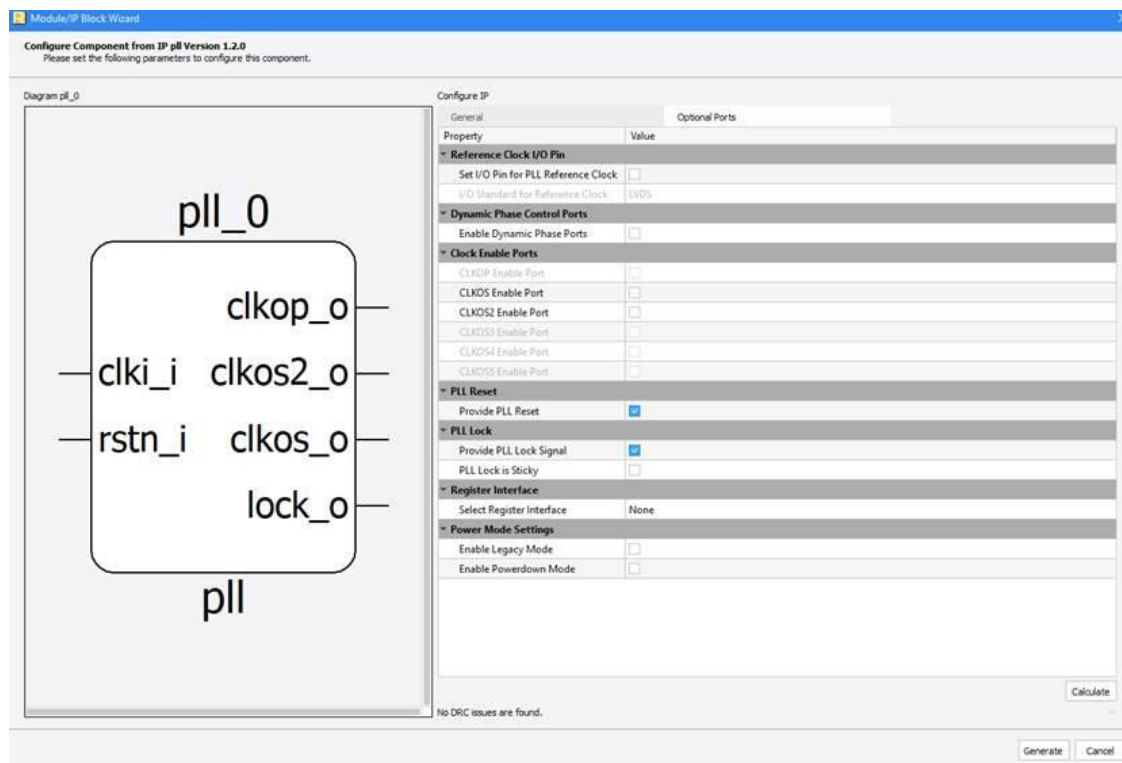


Figure 14.17. Nexus PLL Optional Ports Configuration Tab

Table 14.7. Tab 2, PLL Optional Ports, IP Catalog User Interface

| User Parameters | Range | Default | Description |
|-------------------------------------|-------------------------------------------------------------------------------------------------|-----------|----------------------------------------------------|
| Reference Clock I/O Pin | | | |
| Set I/O Pin for PLL Reference Clock | Checked, Unchecked | Unchecked | Enable/Disable I/O Pin option for reference clock. |
| I/O Standard for Reference Clock | LVDS, SUBLVDS, SLVS, HSTL15_I, HSTL15D_I, LVTTTL33, LVC MOS33, LVC MOS25, LVC MOS18, LVC MOS18H | LVDS | Select type of I/O pin. |
| Dynamic Phase Control Ports | | | |
| Enable Dynamic Phase Ports | Checked, Unchecked | Unchecked | Enable/Disable dynamic phase control ports. |
| Clock Enable Ports | | | |
| CLKOP/CLKOS[n] Enable Port | Checked, Unchecked | Unchecked | Set to provide clock enable port. |
| PLL Reset | | | |
| Provide PLL Reset | Checked, Unchecked | Checked | Set to provide PLL reset port. |
| PLL Lock | | | |
| Provide PLL Lock Signal | Checked, Unchecked | Checked | Set to provide PLL lock port. |
| PLL Lock is Sticky | Checked, Unchecked | Unchecked | Set the behaviour of PLL lock signal. |
| Register Interface | | | |
| Select Register Interface | None, APB, LMMI | None | Select type of register interface. |

| User Parameters | Range | Default | Description |
|----------------------------|--------------------|-----------|---------------------------------|
| Power Mode Settings | | | |
| Enable Legacy Mode | Checked, Unchecked | Unchecked | Set to provide legacy port. |
| Enable Powerdown Mode | Checked, Unchecked | Unchecked | Set to provide power down port. |

For the PLL, IP Catalog sets attributes in the HDL module that are specific to the data rate selected. Although these attributes can be easily changed, they should only be modified by re-running the user interface so that the performance of the PLL is maintained. After the MAP stage in the design flow, the FREQUENCY preferences are included in the preference file to automatically constrain the clocks produced by the PLL. For a step-by-step guide to using IP Catalog, refer to the IP Catalog user manual.

Appendix A. Primary Clock Sources and Distribution

Figure A.1 to Figure A.5 show the inputs into the Primary Clock Network through the MIDMUX into the centermux for each device. There are DCC components at the input of the centermux to allow you to stop the clock to save power.

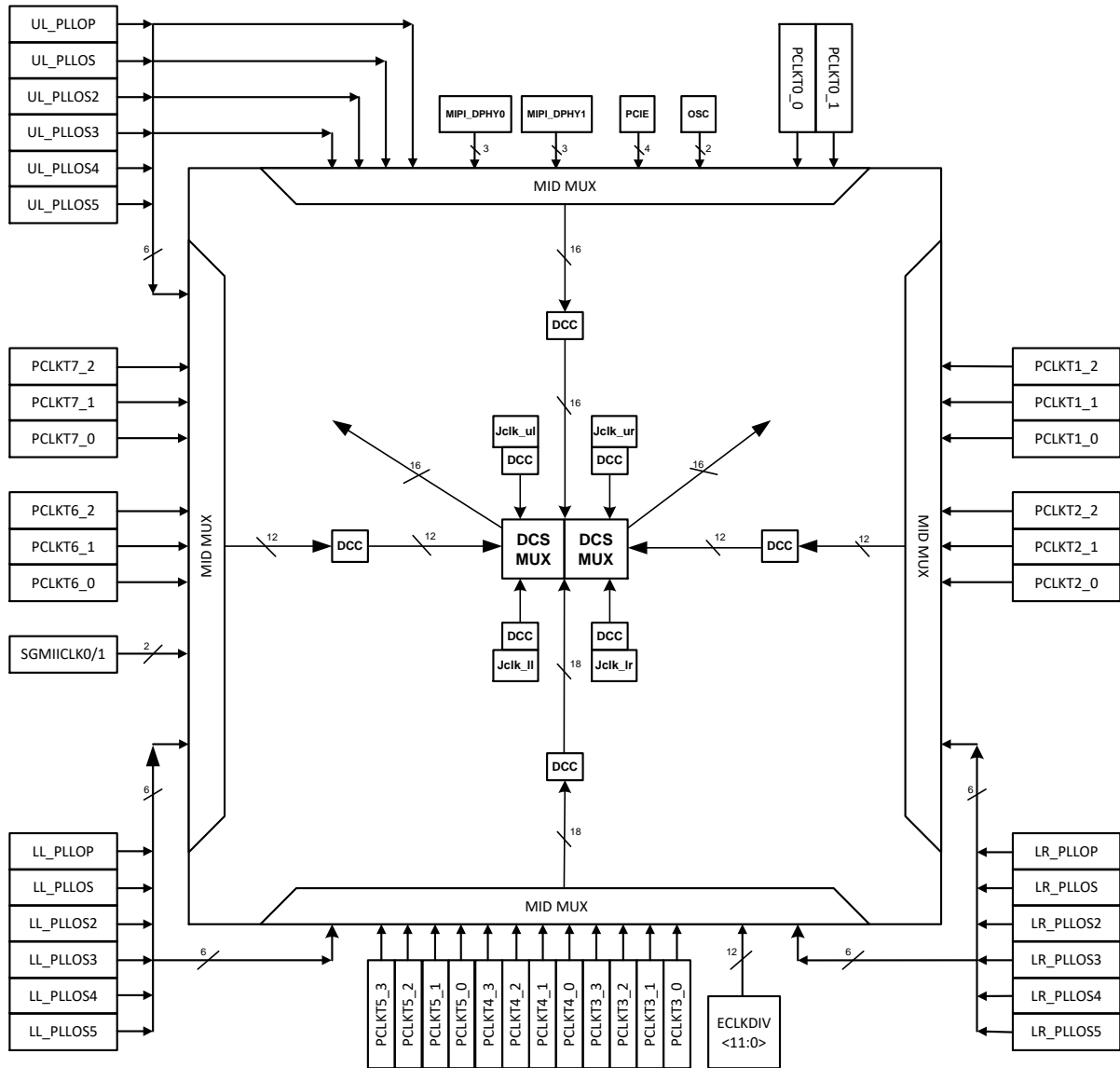


Figure A.1. Nexus Primary Clock Sources and Distribution, LIFCL-40 and LFD2NX-40 Devices

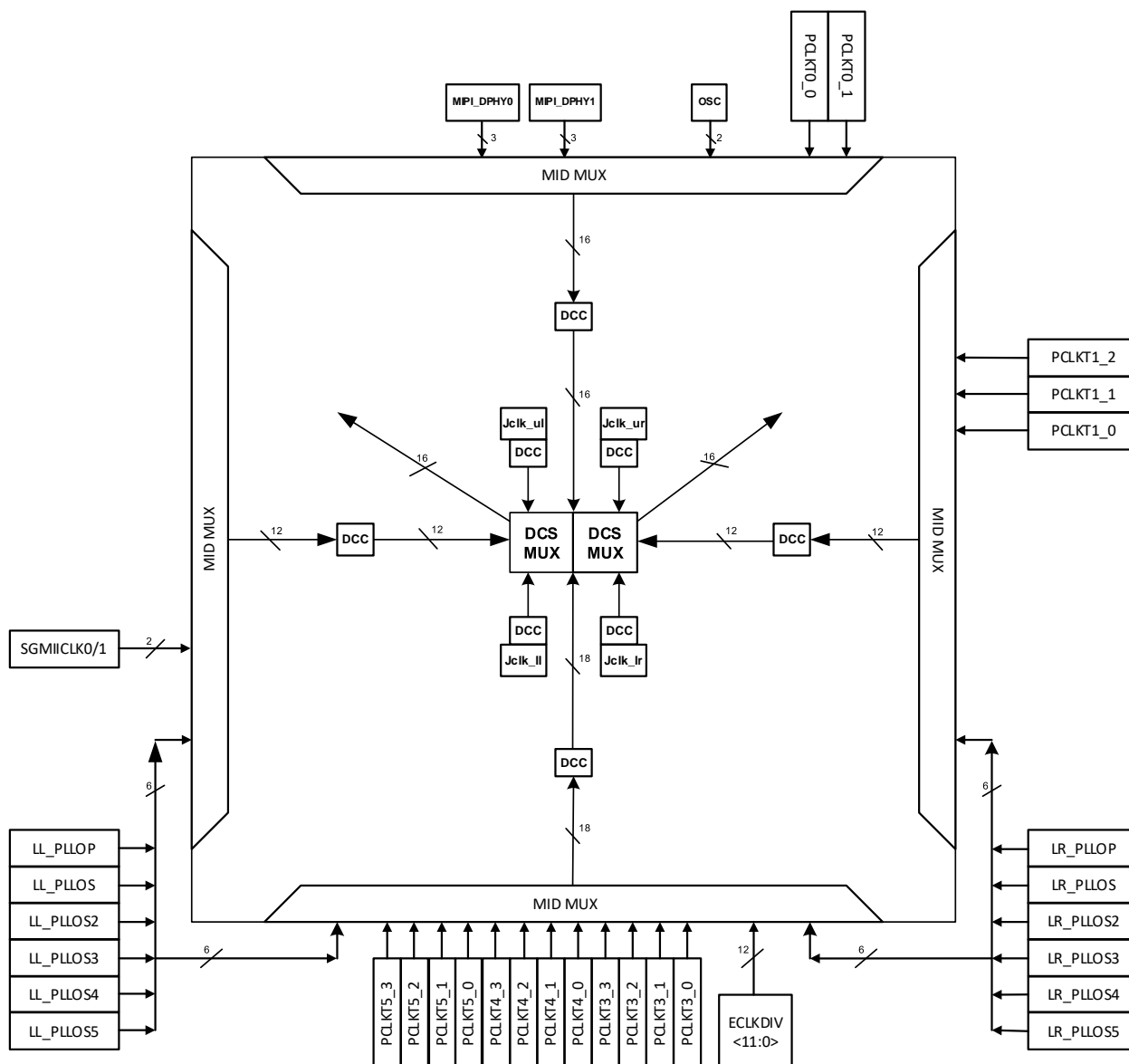


Figure A.2. Nexus Primary Clock Sources and Distribution, LIFCL-17 and LFD2NX-17 Devices

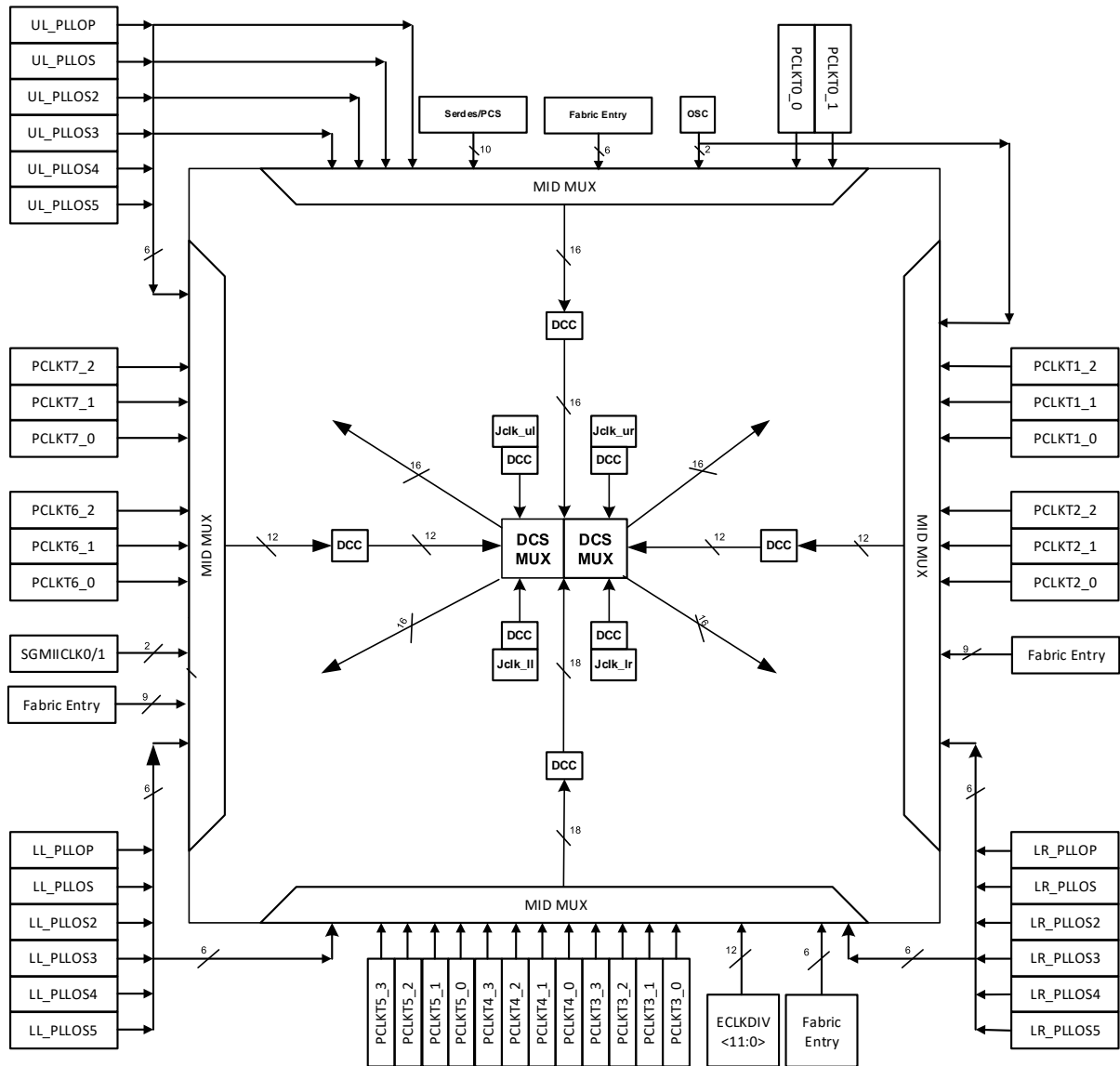
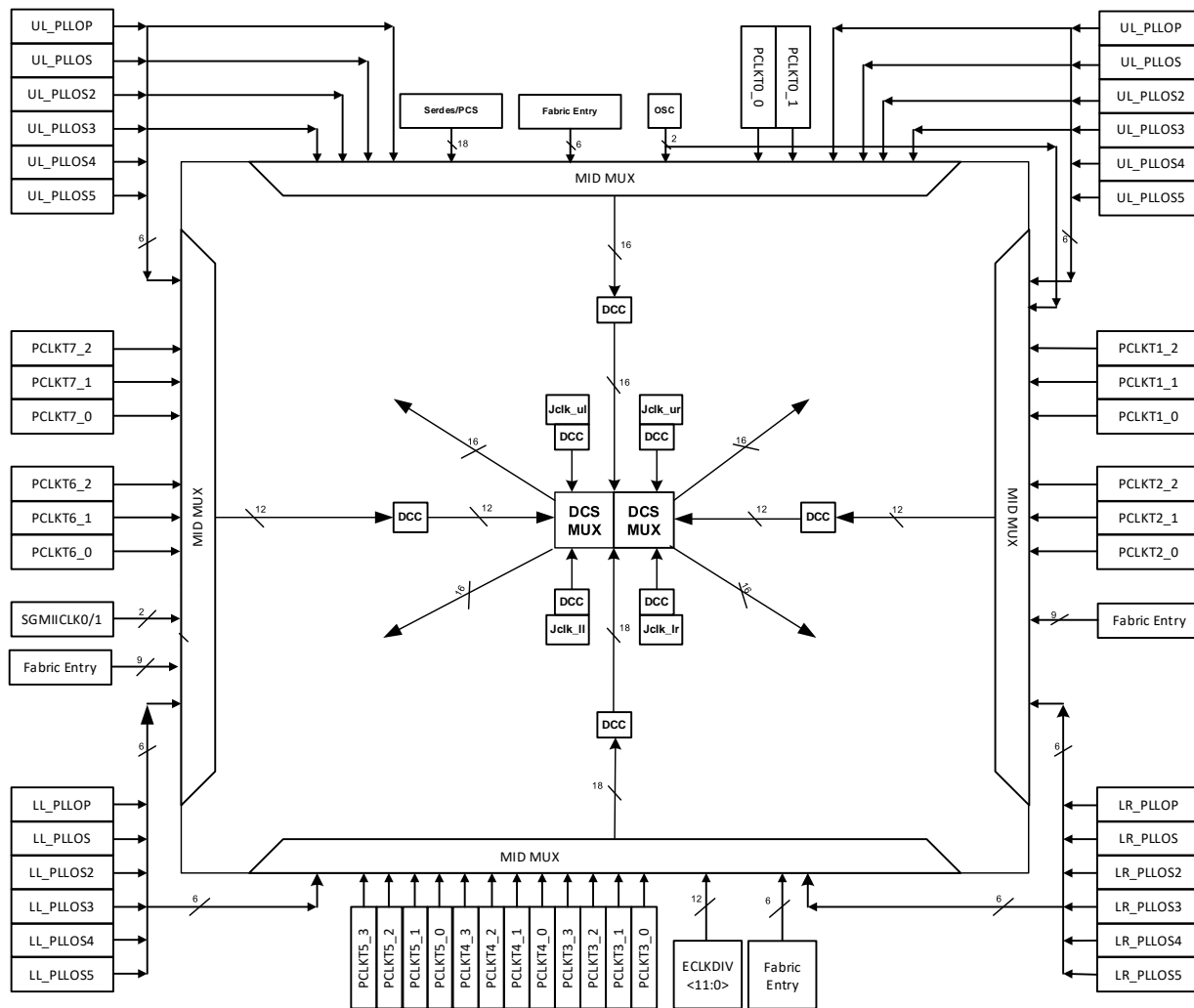


Figure A.3. Nexus Primary Clock Sources and Distribution, LFCPNX-50 Devices



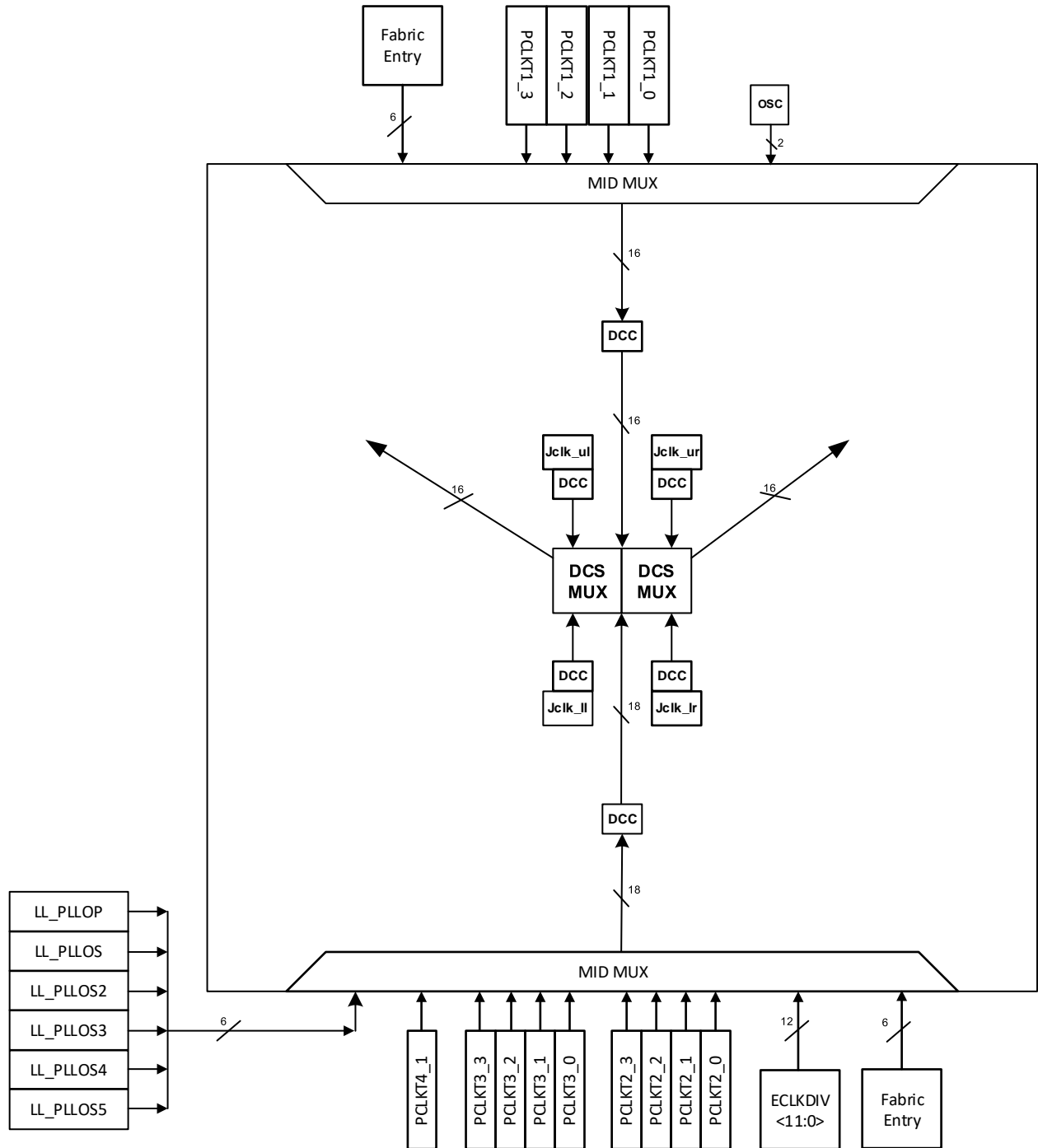


Figure A.5. Nexus Primary Clock Sources and Distribution, LIFCL-33 Devices

Appendix B. Pinout Rules for Clocking in Nexus Devices

In the Nexus device, as with all other architectures, there are general rules and guidelines for board designers to follow. These rules give the best possible timing and allow for a successful design.

In the .csv file where pins are listed, under the *Dual Function* section, you can see the PCLK and PLL input pins listed as below:

Primary Clock Input Pin — PCLK<T/C><Bank>_<0/1/2/3>

Dedicated PLL Input Pin — <LOC>_GPLLO<T/C>_IN

Table B.1. Clock Input Selection Table

| Clock Input | Pin to Use | Clock Routing Resource |
|---------------------------------|----------------|---------------------------------------------------------------|
| Clock Input to Logic Directly | PCLK Input Pin | Uses Primary Clock Routing for the Clock. |
| Clock Input to PLL Only | PLL Input Pin | Uses a Dedicated PLL Input. No Primary Clock Routing is used. |
| Clock Input to Logic and PLL | PCLK Input Pin | Uses Primary Clock Routing for the Clock. |
| Clock input to more than 2 PLLs | PCLK Input Pin | Uses Primary Clock Routing for the Clock. |

Appendix C. PLL LMMI Operation

The Nexus PLL operating parameters can be changed dynamically through the LMMI bus or APB bus. This section uses LMMI nomenclature. All addresses and bit definitions in [PLL Architecture](#) and [LMMI Register Map](#) sections are used identically for APB interface applications. A hard-wired LMMI Bus is used to communicate between the LMMI host and the PLL. See [Lattice Memory Mapped Interface \(LMMI\) and Lattice Interrupt Interface \(LINTR\) User Guide \(FPGA-UG-02039\)](#) for more information about the LMMI bus.

The LMMI Bus on the PLL module provides support for functional operation and simulation. You must connect the LMMI Bus to the LMMI host in their HDL design to make the operand and simulation working properly. The LMMI Bus ports and the corresponding LMMI connections are listed in [Table C.1](#).

Table C.1. PLL Data Bus Port Definition

| PLL Port Name | I/O | Description |
|--------------------|-----|-------------------------------------------------------------------------|
| lmmi_clk_i | I | LMMI clock. |
| lmmi_resetrn_i | I | LMMI reset signal (Active Low). Only reset the bus, not register value. |
| lmmi_offset_i[6:0] | I | LMMI offset address. |
| lmmi_wr_rdn_i | I | LMMI WR/RD signal (write-high/read-low). |
| lmmi_request_i | I | LMMI request signal. |
| lmmi_wdata_i[7:0] | I | LMMI write data. |
| lmmi_ready_o | O | LMMI ready signal. |
| lmmi_rdata_o[7:0] | O | LMMI read data. |
| lmmi_rdata_valid | O | LMMI read data valid signal. |

PLL Architecture

The Nexus PLL has six output sections with flexible configuration settings to support a variety of different applications. IP Catalog is able to support most of the common PLL configurations, but for those users with more complex needs the LMMI bus can be used to change the PLL configuration, which allows for more advanced support options.

Each of the six PLL output sections have similar configuration options. Each output section is assigned a letter designator; A for the CLKOP output, B for the CLKOS output, C for the CLKOS2 output, D for the CLKOS3 output, E for the CLKOS4 output, and F for the CLKOS5 output section.

All LMMI addressable PLL Registers defined in [Table C.2](#) have corresponding shadow registers. The output of the shadow register bits controls the PLL hard IP. To alter the PLL setting through the LMMI Registers, you should write the desired new setting to LMMI Registers when the *Shadow_Reg_Update* bit is at 0. After finishing writing all desired LMMI Registers, write 1 to the *Shadow_Reg_Update* bit, so the new setting takes effect at same time.

LMMI Register Map

The LMMI register map for the PLL registers is shown in [Table C.2](#). The items shaded in grey in [Table C.2](#) and [Table C.3](#) are read only.

Table C.2. LMMI Offset Address Locations for PLL Registers

| Addr | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |
|------|------------------------|--------------------|-------|-------|-------------|-----------------|-------------------|-------------------|
| 00 | lmmi_reserved0[6:0] | | | | | | | Shadow_Reg_Update |
| 01 | lmmi_rotate | lmmi_dyn_sel [2:0] | | | lmmi-sleep | lmmi_stdby | lmmi_pllreset_ena | lmmi_pllpd_n |
| 02 | lmmi_enable_clk [4:0] | | | | | lmmi_dyn_source | lmmi_load_reg | lmmi_direction |
| 03 | lmmi_fbk_cur_ble [3:0] | | | | lmmi_legacy | lmmi_refin_re | lmmi_enable | lmmi_enable |

| Addr | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |
|------|----------------------------|-----------------------------|---------------------------------|-------------------------|------------------------|----------------------------|--------------------------|--------------------|
| | | | | | | set | sync | clk[5] |
| 04 | lmmi_fbk_mask[0] | lmmi_fbk_if_timing_ctl[1:0] | | lmmi_fbk_edge_sel | lmmi_fbk_cur_ble[7:4] | | | |
| 05 | lmmi_fbk_mmd_dig[0] | lmmi_fbk_mask[7:1] | | | | | | |
| 06 | lmmi_fbk_mmd_plus_ctl[0] | lmmi_fbk_mmd_dig[7:1] | | | | | | |
| 07 | lmmi_fbk_pi_rc[1:0] | | lmmi_fbk_pi_bypass | lmmi_fbk_mode[1:0] | | lmmi_fbk_mmd_plus_ctl[3:1] | | |
| 08 | lmmi_fbk_pr_ic[1:0] | | lmmi_fbk_pr_cc[3:0] | | | | lmmi_fbk_pi_rc[3:2] | |
| 09 | lmmi_fbk_rsv[5:0] | | | | | | lmmi_fbk_pr_ic[3:2] | |
| 0A | lmmi_fbk_rsv[13:6] | | | | | | | |
| 0B | lmmi_ref_mask[3:0] | | | | lmmi_ref_integer_mode | lmmi_fbk_integer_mode | lmmi_fbk_rsv[15:14] | |
| 0C | lmmi_ref_mmd_dig[3:0] | | | | lmmi_ref_mask[7:4] | | | |
| 0D | lmmi_ref_mmd_in[3:0] | | | | lmmi_ref_mmd_dig[7:4] | | | |
| 0E | lmmi_ref_mmd_plus_ctl[3:0] | | | | lmmi_ref_mmd_in[7:4] | | | |
| 0F | lmmi_ldt_int_lock_sticky | lmmi_ldt_lock_sel[2:0] | | | lmmi_ldt_lock[1:0] | | lmmi_ref_timing_ctl[1:0] | |
| 10 | lmmi_ssc_delta[3:0] | | | | lmmi_flock_src_sel | lmmi_flock_en | lmmi_flock_ctrl[1:0] | |
| 11 | lmmi_ssc_delta[11:4] | | | | | | | |
| 12 | lmmi_ssc_en_sdm | lmmi_ssc_en_center_in | lmmi_ssc_dither | lmmi_ssc_delta_ctl[1:0] | | lmmi_ssc_delta[14:12] | | |
| 13 | lmmi_ssc_f_code[6:0] | | | | | | | lmmi_ssc_en_ssc |
| 14 | lmmi_ssc_f_code[14:7] | | | | | | | |
| 15 | lmmi_ssc_n_code[7:0] | | | | | | | |
| 16 | lmmi_ssc_step_in[0] | lmmi_ssc_square_mode | lmmi_ssc_reg_weighting_sel[2:0] | | | lmmi_ssc_pi_bypass | lmmi_ssc_order | lmmi_ssc_n_code[8] |
| 17 | lmmi_ssc_tbase[1:0] | | lmmi_ssc_step_in[6:1] | | | | | |
| 18 | lmmi_ssc_tbase[9:2] | | | | | | | |
| 19 | lmmi_bw_ctl_bias[0] | lmmi_delay_ctrl | lmmi_float_cp | lmmi_i_ctrl[2:0] | | | lmmi_ssc_tbase[11:10] | |
| 1A | lmmi_ipi_cmpn[2:0] | | | Not Used | Not Used | lmmi_bw_ctl_bias[3:1] | | |
| 1B | lmmi_ipp_sel[2:0] | | | lmmi_ipp_ctrl[3:0] | | | | lmmi_ipi_cmpn[3] |
| 1C | lmmi_v2i_pp_ictrl[0] | lmmi_v2i_1v_en | lmmi_v2i_kvco_sel[3:0] | | | | lmmi_fast_lock_en | lmmi_ipp_sel[3] |
| 1D | lmmi_openloop_en | lmmi_v2i_pp_res[2:0] | | | lmmi_v2i_pp_ictrl[4:1] | | | |
| 1E | lmmi_cripple[2:0] | | | lmmi_cset[3:0] | | | | lmmi_reset_lf |
| 1F | lmmi_mfg_ctrl[2:0] | | | lmmi_kp_vco[4:0] | | | | |
| 20 | lmmi_mfg_en | lmmi_force_filter | lmmi_ipi_comp_en | lmmi_ipi_cmp[3:0] | | | | lmmi_mfg_ctrl[3] |
| 21 | lmmi_phib[1:0] | | lmmi_phia[2:0] | | | lmmi_mfg_sel[2:0] | | |
| 22 | lmmi_phie[0] | lmmi_phid[2:0] | | | lmmi_phic[2:0] | | | lmmi_phib[2] |
| 23 | lmmi_sel_outc | lmmi_sel_outb | lmmi_sel_outa | lmmi_phif[2:0] | | | lmmi_phie[2:1] | |
| 24 | lmmi_dela[4:0] | | | | | lmmi_sel_outf | lmmi_sel_oute | lmmi_sel_outd |

| Addr | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |
|------|----------------------------|----------------------|-------------------------|-----------------------|----------------|-----------------------|-----------------------|--------------------------|
| 25 | lmmi_delb[5:0] | | | | | | lmmi_dela[6:5] | |
| 26 | lmmi_delc[6:0] | | | | | | | lmmi_delb[6] |
| 27 | lmmi_dele[0] | lmmi_deld[6:0] | | | | | | |
| 28 | lmmi_delf[1:0] | | lmmi_dele[6:1] | | | | | |
| 29 | lmmi_diva[2:0] | | | lmmi_delf[6:2] | | | | |
| 2A | lmmi_divb[3:0] | | | | lmmi_diva[6:3] | | | |
| 2B | lmmi_divc[4:0] | | | | | lmmi_divb[6:4] | | |
| 2C | lmmi_divd[5:0] | | | | | | lmmi_divc[6:5] | |
| 2D | lmmi_dive[6:0] | | | | | | | lmmi_divd[6] |
| 2E | lmmi_mfgout1_sel[0] | lmmi_divf[6:0] | | | | | | |
| 2F | lmmi_clkop_trim[2:0] | | | lmmi_mfgout2_sel[2:0] | | | lmmi_mfgout1_sel[2:1] | |
| 30 | lmmi_clkos2_trim[2:0] | | | lmmi_clkos_trim[3:0] | | | | lmmi_clkop_trim[3] |
| 31 | lmmi_clkos4_trim[2:0] | | | lmmi_clkos3_trim[3:0] | | | | lmmi_clkos2_trim[3] |
| 32 | lmmi_trimos2_bypass_n | lmmi_trimos_bypass_n | lmmi_trimop_bypass_n | lmmi_clkos5_trim[3:0] | | | | lmmi_clkos4_trim[3] |
| 33 | lmmi_div_del[3:0] | | | | lmmi_pllpdn_en | lmmi_trimos5_bypass_n | lmmi_trimos4_bypass_n | lmmi_trimos3_bypass_n |
| 34 | lmmi_phase_sel_del_p1[1:0] | | lmmi_phase_sel_del[2:0] | | | lmmi_div_del[6:4] | | |
| 35 | lmmi_reserved[6:0] | | | | | | | lmmi_phase_sel_del_p1[2] |

Table C.3. PLL Registers Descriptions

| Register Name | Register Addr (Hex) | Size (Bits) | Description | Default Value | User Access ¹ |
|---------------------|---------------------|-------------|---------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------|---------------|--------------------------|
| lmmi_reserved0[6:0] | 00[7:1] | 6 | Reserved | 7'b0000000 | RO |
| Shadow_Reg_Update | 00[0] | 1 | 1'b0: Allows LMMI register bit value updates. 1'b1: Set to 1 to transfer the updated LMMI register content to the active shadow registers. All changed register bit settings become active simultaneously. | 1'b0 | R/W |
| lmmi_pllpd_n | 01[0] | 1 | 1'b0 – PLL is not used. | 1'b0 | R/W |
| lmmi_pllreset_ena | 01[1] | 1 | Active HIGH; Enable PLLRESET CIB signal | 1'b0 | R/W |
| lmmi_stdby | 01[2] | 1 | NOT Supported. Must be set to 1'b0. | 1'b0 | RO |
| lmmi_sleep | 01[3] | 1 | 1'b1 – enable PLL to support sleep/stop mode | 1'b0 | R/W |
| lmmi_dyn_sel[2:0] | 01[6:4] | 3 | Output clock phase shift selection, only one output is shifted at one time. 000 – CLKOS 001 – CLKOS2 010 – CLKOS3 011 – CLKOS4 100 – CLKOS5 101 – CLKOP | 3'b000 | R/W |

| Register Name | Register Addr (Hex) | Size (Bits) | Description | Default Value | User Access 1 |
|-----------------------------|---------------------|-------------|-------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------|---------------|---------------|
| Immi_rotate | 01[7] | 1 | LMMI equivalent of CIB rotate signal. Valid if Immi_dyn_source = 0. Initiate a change from current VCO clock phase to an earlier or later phase on the negative edge of Immi_rotate. | 1'b0 | R/W |
| Immi_direction | 02[0] | 1 | LMMI equivalent of CIB direction signal. Valid if Immi_dyn_source = 0. Specify direction that Immi_rotate changes VCO phase. 0 – Phase rotates to later phase. 1 – Phase rotates to earlier phase. | 1'b0 | R/W |
| Immi_load_reg | 02[1] | 1 | LMMI equivalent of CIB load signal. Valid if Immi_dyn_source = 0. Initiate a divider output phase shift on negative edge of LMMI_LOAD_REG. After two consecutive output clock cycles, one divider cycle use Immi_del[A/B/C/D/E/F] instead of Immi_div[A/B/C/D/E/F]. Only the output which is addressed by LMMI_DYN_SEL is shifted. On the falling edge of the LMMI_LOAD_REG, the internal phase shift starts. The minimum pulse width is 10 ns. The setting control signals need to have 5 ns setup time with respect to the LMMI_LOAD_REG falling edge. | 1'b0 | R/W |
| Immi_dyn_source | 02[2] | 1 | 1'b0 select LMMI signals for dynamic phase shift. 1'b1 select CIB signals for dynamic phase shift. | 1'b0 | R/W |
| Immi_enable_clk[5:0] | 02[7:3] 03[0] | 6 | 1'b1 enables corresponding output mapped as <clkos5, clkos4, clkos3, clkos2, clkos, clkop>. | 6'b000000 | R/W |
| Immi_enable_sync | 03[1] | 1 | Active HIGH; Enable synchronous disable/enable of secondary clocks CLKOS, CLKOS2, CLKOS3, CLKOS4, CLKOS5 with respect to CLKOP. | 1'b0 | R/W |
| Immi_refin_reset | 03[2] | 1 | Active High. Enable PLL internal reset generated after reference mux dynamic selection from CIB. | 1'b0 | R/W |
| Immi_legacy | 03[3] | 1 | Active HIGH; enable Legacy mode. | 1'b0 | R/W |
| Immi_fbk_cur_ble[7:0] | 03[7:4] 04[3:0] | 8 | Bleeding current for PI to adjust the linearity. | 8'H00 | RO |
| Immi_fbk_edge_sel | 04[4] | 1 | Select the positive or negative phase of PI output. 0: positive phase. 1: negative phase. | 1'b0 | RO |
| Immi_fbk_if_timing_ctl[1:0] | 04[6:5] | 2 | Interface timing control for feedback divider. | 1'b0 | RO |
| Immi_fbk_mask[7:0] | 04[7] 05[6:0] | 8 | Minimum divider ratio control word for feedback divider. For example, if n_pll<7:0> or mmd_dig<7:0> is less than Immi_fbk_mask<7:0>, then the MMD divider ratio is determined by Immi_fbk_mask<7:0>. Otherwise, the divider ratio is determined by n_pll<7:0> or mmd_dig<7:0>. | 8'b00001000 | RO |
| Immi_fbk_mmd_dig[7:0] | 05[7] 06[6:0] | 8 | MMD divider ratio setting in integer mode | 8'b00001000 | RO |
| Immi_fbk_mmd_puls_ctl[3:0] | 06[7] 07[2:0] | 4 | Pulse width control for MMD output clock. If Immi_fbk_mmd_puls_ctl<3:0>=4'b0110, it means that there's 6 VCO cycles in the MMD output clock. If divider value > 2, 4'b0001. | 4'b0000 | RO |
| Immi_fbk_mode[1:0] | 07[4:3] | 2 | Reserved floating control bits. | 2'b00 | RO |
| Immi_fbk_pi_bypass | 07[5] | 1 | PI bypass control bit. It should be same as Immi_ssc_pi_bypass. 0: PI not bypass. 1: PI bypass; | 1'b0 | R/W |

| Register Name | Register Addr (Hex) | Size (Bits) | Description | Default Value | User Access 1 |
|----------------------------|--------------------------------|-------------|--------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------|---------------|---------------|
| Immi_fbk_pi_rc[3:0] | 07[7:6] 08[1:0] | 4 | RC time constant control in PI. | 4'b1100 | RO |
| Immi_fbk_pr_cc<3:0> | 08[5:2] | 4 | Current control for PI to adjust the linearity. | 4'b0000 | RO |
| Immi_fbk_pr_ic<3:0> | 08[7:6] 09[1:0] | 4 | Bias current control for PI. | 4'b1000 | RO |
| Immi_fbk_rsv[15:0] | 09[7:2] 0A[7:0] 0B[[1:0] | 16 | Reserved control bit for feedback divider. | 16'H0000 | RO |
| Immi_fbk_integer_mode | 0B[2] | 1 | Enable the integer mode for feedback divider. | 1'b0 | R/W |
| Immi_ref_integer_mode | 0B[3] | 1 | Integer mode control bit for reference clock pre-divider | 1'b0 | R/W |
| Immi_ref_mask[7:0] | 0B[7:4] 0C[3:0] | 8 | Minimum divider ratio control word for reference pre-divider. | 8'H00 | R/W |
| Immi_ref_mmd_dig[7:0] | 0C[7:4] 0D[3:0] | 8 | MMD divider ratio setting for reference pre-divider when Immi_ref_integer_mode=1 | 8'H08 | R/W |
| Immi_ref_mmd_in[7:0] | 0D[7:4] 0E[3:0] | 8 | MMD divider ratio setting for reference pre-divider when Immi_ref_integer_mode=0. | 8'H08 | R/W |
| Immi_ref_mmd_puls_ctl[3:0] | 0E[7:4] | 4 | Pulse width control for MMD output clock in reference pre-divider. If divider value > 2, 4'b0001. | 4'b0000 | RO |
| Immi_ref_timing_ctl[1:0] | 0F[1:0] | 2 | Interface timing control for reference divider. Default setting is 2'b00. | 2'b00 | RO |
| Immi_ldt_lock[1:0] | 0F[3:2] | 2 | Frequency lock-detector resolution sensitivity 00 = takes about 98304 PFDFBK cycles to lock 01 = takes about 24576 PFDFBK cycles to lock 10 = takes about 6144 PFDFBK cycles to lock 11 = takes about 1536 PFDFBK cycles to lock | 2'b11 | R/W |
| Immi_ldt_lock_sel[2:0] | 0F[6:4] | 3 | Lock-detector type select: 000 = UNSTICKY freq lock ¹ 001 = STICKY phase lock (freq lock first detected then phase lock 1st detected)* 010 = STICKY frequency lock (freq lock first detected) ¹ 011 = UNSTICKY freq and STICKY phase lock ¹ 100 = UNSTICKY freq lock 101 = UNSTICKY phase lock 110 = STICKY freq lock 111 = UNSTICKY freq and STICKY phase lock | 3'b000 | R/W |
| Immi_ldt_int_lock_sticky | 0F[7] | 1 | Active HIGH to have INT_LOCK STICKY. Default to be 1'b1. 1'b0 for PDE/DE purpose. | 1'b1 | Read Only |
| Immi_flock_ctrl[1:0] | 10[1:0] | 2 | 2 bits control the fast lock period. 00 is 1x, 01 is 2x, 10 is 4x, 11 is 8x | 2'b01 | R/W |
| Immi_flock_en | 10[2] | 1 | Active high. To enable fast lock. | 1'b1 | R/W |
| Immi_flock_src_sel | 10[3] | 1 | fast lock source selection: 0 is ref clock. 1 is feedback clock. | 1'b0 | R/W |
| Immi_ssc_delta[14:0] | 10[7:4] 11[7:0] 12[2:0] | 15 | RSVD | 15'H0000 | RO |
| Immi_ssc_delta_ctl[1:0] | 12[4:3] | 2 | RSVD | 2'b00 | RO |
| Immi_ssc_dither | 12[5] | 1 | Dither enable or disable for SDM. 0: disable; 1: enable | 1'b0 | RO |

| Register Name | Register Addr (Hex) | Size (Bits) | Description | Default Value | User Access 1 |
|---------------------------------|-------------------------------|-------------|----------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------|---------------|---------------|
| Immi_ssc_en_center_in | 12[6] | 1 | Down triangle or central triangle control bit in SSC profile generator. 0: down-triangle; 1: central-triangle. | 1'b0 | RO |
| Immi_ssc_en_sdm | 12[7] | 1 | Enable or disable the SDM. 0: disable the SDM. 1: enable the SDM. In stair-wave SSC mode, it should be set to 0. | 1'b0 | RO |
| Immi_ssc_en_ssc | 13[0] | 1 | Enable or disable the SSC profile generator. 0: disable the SSC generator. 1: enable the SSC generator. | 1'b0 | RO |
| Immi_ssc_f_code[14:0] | 13[7:1] 14[7:0] | 15 | Fractional part of the feedback divider ratio. | 15'H0000 | RO |
| Immi_ssc_n_code[8:0] | 15[7:0] 16[0] | 9 | Integer part of feedback divider ratio | 9'b000010100 | RO |
| Immi_ssc_order | 16[1] | 1 | SDM order control bit. 0: SDM order=1; 1: SDM order=2, MASH1-1 | 1'b0 | RO |
| Immi_ssc_pi_bypass | 16[2] | 1 | PI bypass control bit. 0: PI not bypass; 1: PI bypass; it should be same as Immi_fbk_pi_bypass. | 1'b0 | RO |
| Immi_ssc_reg_weighting_sel[2:0] | 16[5:3] | 3 | Weighting control bit for Immi_ssc_step_in<6:0> | 3'b000 | RO |
| Immi_ssc_square_mode | 16[6] | 1 | Two-point FSK modulation control bit. 0: disable; 1: enable 2-point FSK. | 1'b0 | RO |
| Immi_ssc_step_in[6:0] | 16[7] 17[5:0] | 7 | SSC modulation depth control bit. | 7'b0000000 | RO |
| Immi_ssc_tbase[11:0] | 17[7:6] 18[7:0] 19[1:0] | 12 | SSC modulation frequency control. The frequency should be 30~33kHz. | 12'H000 | RO |
| Immi_l_ctrl[2:0] | 19[4:2] | 3 | current tuning: 000:10 μ A; 001:8.3 μ A; 010:14.9 μ A; 011:12.4 μ A; 100:19.8 μ A; 101:17.3 μ A; 110:24.8 μ A; 111:22.3 μ A | 3'b000 | RO |
| Immi_float_cp | 19[5] | 1 | Active HIGH to tri-state the ICP output. | 1'b0 | RO |
| Immi_delay_ctrl | 19[6] | 1 | Control signal to adjust the delay of the PFD; default 1b0=200 ps; 1b1=300 ps | 1'b0 | RO |
| Immi_bw_ctl_bias[3:0] | 19[7] 1A[2:0] | 4 | Input control signal to tune the bias current of ppath cp, When the bit increase, the bias current increase. This current branch is combined with the lvco_fb current. | 4'b0000 | RO |
| Immi_ipi_cmpn[3:0] | 1A[7:5] 1B[0] | 4 | Input control bits to compensate i-path bias current | 4'b0000 | RO |
| Immi_ipp_ctrl[3:0] | 1B[4:1] | 4 | Input control signal to tune the bias current of ppath cp. ipp_ctrl<3:2> was used to control the bias voltage of the cpp_bias; ipp_ctrl<1:0> was used to tune the current of bias, start from 5uA to 20uA with 5uA step from each bit. | 4'b0110 | RO |
| Immi_ipp_sel[3:0] | 1B[7:5] 1C[0] | 4 | Input control signal to select which ppath cp is on, there are 4 branches at max p-path current function: [5uA+ipp_ctrl<1:0>x5uA]/3xbw_ctl_bias<3:0>x[ipp_sel<3>+ipp_sel<2>+ipp_sel<1>+ipp_sel<0>] | 4'b1111 | RO |
| Immi_fast_lock_en | 1C[1] | 1 | Enable signal for fast lock, default to 1'b1 | 1'b0 | RO |

| Register Name | Register Addr (Hex) | Size (Bits) | Description | Default Value | User Access 1 |
|------------------------|---------------------|-------------|---------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------|--------------------------------------------------|---------------|
| Immi_v2i_kvco_sel[3:0] | 1C[5:2] | 4 | v2i kvco slope control, $10 + \text{Dec}(\text{Immi_v2i_kvco_sel}) \times 5$ | 4'b1001 (for 0.9V) 4'b0100 (for 1V) | RO |
| Immi_v2i_1v_en | 1C[6] | 1 | 1 V supply enable or disable. 0: disable. 1: enable | 1'b0 (for 0.9v) 1'b1 (for 1v) | RO |
| Immi_v2i_pp_ictrl<4:0> | 1C[7] 1D[3:0] | 5 | P-path v2i gm control | 5'b00110 | RO |
| Immi_v2i_pp_res<2:0> | 1D[6:4] | 3 | P-path high frequency pole resistor control 000: 11.3K 001: 11K 010:10.7K 011:10.3K 100: 10K 101:9.7K 110: 9.3K 111:9K | 3'b000 | RO |
| Immi_openloop_en | 1D[7] | 1 | open loop mode enable for mfg testing. | 1'b0 | RO |
| Immi_reset_lf | 1E[0] | 1 | lpf reset enable, default to 1'b0 | 1'b0 | RO |
| Immi_cset[3:0] | 1E[4:1] | 4 | LPF cap control 0000:8p; 0001:12p 0010:16p 0011:20p 0100:24p 0101:28p 0110:32p 0111:36p 1000:40p 1001:44p 1010:48p 1011:52p 1100:56p 1101:60p 1110:64p 1111:68p | 4'b1000 | RO |
| Immi_cripple[2:0] | 1E[7:5] | 3 | LPF cap control 000:1p; 001:3p 010:5p 011:7p 100:9p 101:11p 110:13p 111:15p | 3'b010 | RO |
| Immi_kp_vco[4:0] | 1F[4:0] | 5 | $90 + \text{Dec}(\text{Immi_kp_vco}) \times 10;$ | 5'b11001 (for 0.9V) 5'b00011 (for 1.0v) | RO |
| Immi_mfg_ctrl[3:0] | 1F[7:5] 20[0] | 4 | mfg internal vctrl selection 0000:0; 1111:vdd, step is 55 mV | 4'b0000 | RO |
| Immi_ipi_cmp[3:0] | 20[4:1] | 4 | i-path CP compensate up/dn mismatch at process variation | 4'b1000 | RO |
| Immi_ipi_cmp_en | 20[5] | 1 | Enable ipi_cmp combine with Immi_en_ipi_cmp to adjust the delay of the PFD, this bit is LSB; 2'b00=160ps; 2'b10=200ps; 2'b11=230ps; 2'b11=300ps | 1'b0 | RO |
| Immi_force_filter | 20[6] | 1 | force internal vctrl=analog pad | 1'b0 | RO |
| Immi_mfg_en | 20[7] | 1 | mfg feature enable pin | 1'b0 | RO |
| Immi_mfg_sel[2:0] | 21[2:0] | 3 | mfg current mux selection 000:I path CP up current; 001:P path CP up current; 010:I path V2I current 011:P path V2I current 100:I path CP dn current 101:P path CP dn current 110:NA 111:NA | 3'b000 | RO |
| Immi_phiA [2:0] | 21[5:3] | 3 | Select VCO phase-shift (0..7) for A section | 3'b000 | R/W |
| Immi_phiB [2:0] | 21[7:6] 22[0] | 3 | Select VCO phase-shift (0..7) for B section | 3'b000 | R/W |
| Immi_phiC [2:0] | 22[3:1] | 3 | Select VCO phase-shift (0..7) for C section | 3'b000 | R/W |

| Register Name | Register Addr (Hex) | Size (Bits) | Description | Default Value | User Access 1 |
|-----------------------|---------------------|-------------|--------------------------------------------------------------------------------------------------------------------------------------|---------------|---------------|
| Immi_phid [2:0] | 22[6:4] | 3 | Select VCO phase-shift (0..7) for D section | 3'b000 | R/W |
| Immi_phie [2:0] | 22[7] 23[1:0] | 3 | Select VCO phase-shift (0..7) for E section | 3'b000 | R/W |
| Immi_phif [2:0] | 23[4:2] | 3 | Select VCO phase-shift (0..7) for F section | 3'b000 | R/W |
| Immi_sel_outa | 23[5] | 1 | Select output to CLKOP | 1'b0 | R/W |
| Immi_sel_outb | 23[6] | 1 | Select output to CLKOS | 1'b0 | R/W |
| Immi_sel_outc | 23[7] | 1 | Select output to CLKOS2 | 1'b0 | R/W |
| Immi_sel_outd | 24[0] | 1 | Select output to CLKOS3 | 1'b0 | R/W |
| Immi_sel_oute | 24[1] | 1 | Select output to CLKOS4 | 1'b0 | R/W |
| Immi_sel_outf | 24[2] | 1 | Select output to CLKOS5 | 1'b0 | R/W |
| Immi_dela[6:0] | 24[7:3] 25[1:0] | 7 | Delay A section output DELA VCO clock cycles with respect to VCO phase 0, or REFBUF if in VCO bypass mode (post-divider phase-shift) | 7'H00 | R/W |
| Immi_delb[6:0] | 25[7:2] 26[0] | 7 | Delay B section output DELB VCO clock cycles with respect to VCO phase 0, or REFBUF if in VCO bypass mode (post-divider phase-shift) | 7'H00 | R/W |
| Immi_delc[6:0] | 26[7:1] | 7 | Delay C section output DELC VCO clock cycles with respect to VCO phase 0, or REFBUF if in VCO bypass mode (post-divider phase-shift) | 7'H00 | R/W |
| Immi_deld[6:0] | 27[6:0] | 7 | Delay D section output DELD VCO clock cycles with respect to VCO phase 0, or REFBUF if in VCO bypass mode (post-divider phase-shift) | 7'H00 | R/W |
| Immi_dele[6:0] | 27[1] 28[5:0] | 7 | Delay E section output DELE VCO clock cycles with respect to VCO phase 0, or REFBUF if in VCO bypass mode (post-divider phase-shift) | 7'H00 | R/W |
| Immi_delf[6:0] | 28[7:6] 29[4:0] | 7 | Delay F section output DELF VCO clock cycles with respect to VCO phase 0, or REFBUF if in VCO bypass mode (post-divider phase-shift) | 7'H00 | R/W |
| Immi_diva [6:0] | 29[7:5] 2A[3:0] | 7 | Output dividers setting for clkop: divide value = DIVA + 1 | 7'H00 | R/W |
| Immi_divb [6:0] | 2A[7:4] 2B[2:0] | 7 | Output dividers setting for clkos: divide value = DIVB + 1 | 7'H00 | R/W |
| Immi_divc [6:0] | 2B[7:3] 2C[1:0] | 7 | Output dividers setting for clkos2: divide value = DIVC + 1 | 7'H00 | R/W |
| Immi_divd [6:0] | 2C[7:2] 2D[0] | 7 | Output dividers setting for clkos3: divide value = DIVD + 1 | 7'H00 | R/W |
| Immi_dive[6:0] | 2D[7:1] | 7 | Output dividers setting for clkos4: divide value = DIVE + 1 | 7'H00 | R/W |
| Immi_divf[6:0] | 2E[6:0] | 7 | Output dividers setting for clkos5: divide value = DIVF + 1 | 7'H00 | R/W |
| Immi_mfgout1_sel[2:0] | 2E[7] 2F[1:0] | 3 | PLL_MFGOUT1 selection bits | 3'b000 | RO |
| Immi_mfgout2_sel[2:0] | 2F[4:2] | 3 | PLL_MFGOUT2 selection bits | 3'b000 | RO |
| Immi_clkop_trim[3:0] | 2F[7:5] 30[0] | 4 | CLKOP output edge trim | 4'b0000 | RO |
| Immi_clkos_trim[3:0] | 30[4:1] | 4 | CLKOS output edge trim | 4'b0000 | RO |
| Immi_clkos2_trim[3:0] | 30[7:5] 31[0] | 4 | CLKOS2 output edge trim | 4'b0000 | RO |
| Immi_clkos3_trim[3:0] | 31[4:1] | 4 | CLKOS3 output edge trim | 4'b0000 | RO |
| Immi_clkos4_trim[3:0] | 31[7:5] 32[0] | 4 | CLKOS4 output edge trim | 4'b0000 | RO |

| Register Name | Register Addr (Hex) | Size (Bits) | Description | Default Value | User Access ¹ |
|----------------------------|---------------------|-------------|-------------------------------------------|---------------|--------------------------|
| lmmi_clkos5_trim[3:0] | 32[4:1] | 4 | CLKOS5 output edge trim | 4'b0000 | RO |
| lmmi_trimop_bypass_n | 32[5] | 1 | CLKOP output edge trim bypass | 1'b0 | RO |
| lmmi_trimos_bypass_n | 32[6] | 1 | CLKOS output edge trim bypass | 1'b0 | RO |
| lmmi_trimos2_bypass_n | 32[7] | 1 | CLKOS2 output edge trim bypass | 1'b0 | RO |
| lmmi_trimos3_bypass_n | 33[0] | 1 | CLKOS3 output edge trim bypass | 1'b0 | RO |
| lmmi_trimos4_bypass_n | 33[1] | 1 | CLKOS4 output edge trim bypass | 1'b0 | RO |
| lmmi_trimos5_bypass_n | 33[2] | 1 | CLKOS5 output edge trim bypass | 1'b0 | RO |
| lmmi_pllpdn_en | 33[3] | 1 | Active high to enable pllpd_n CIB control | 1'b0 | R/W |
| lmmi_div_del[6:0] | 33[7:4] 34[2:0] | 7 | The internal delay path divider | 7'b0000001 | R/W |
| lmmi_phase_sel_del[2:0] | 34[5:3] | 3 | The internal phase delay selection path | 3'b000 | R/W |
| lmmi_phase_sel_del_p1[2:0] | 34[7:6] 35[0] | 3 | The internal phase delay selection path1 | 3'b000 | R/W |
| lmmi_reserved[6:0] | 35[7:1] | 7 | Reserved | 7'b0000000 | RO |

Notes:

1. Gated with pll_wakeup_sync pin.
2. R/W = Read and Write; RO = Read Only
3. ppath (p-path) stands for proportional path, I-path stands for Integral path which is a dual tuning PLL using PI tuning loop, and CP is abbreviated for Charge Pump.

Technical Support Assistance

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Revision History

Revision 2.0, June 2022

| Section | Change Summary |
|----------------------------------------------------|----------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------|
| Introduction | Updated Table 1.1. Number of PLLs, Edge Clocks, and Clock Dividers to add LIFCL-33. |
| Nexus Top-Level View | Added Figure 3.4. CrossLink-NX-33 Clocking Structure . |
| sysCLOCK PLL | <ul style="list-style-type: none"> Added information on one PLL for LIFCL-33 in sysCLOCK PLL Overview. Added Figure 14.7. PLL Input Pins for LIFCL-33. |
| Appendix A. Primary Clock Sources and Distribution | Added Figure A.5. Nexus Primary Clock Sources and Distribution, LIFCL-33 Devices . |

Revision 1.9, May 2022

| Section | Change Summary |
|--------------|-------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------|
| All | <ul style="list-style-type: none"> Added MachXO5-NX support across the document. Changed SERDES to SerDes across the document. |
| Introduction | Updated Table 1.1. Number of PLLs, Edge Clocks, and Clock Dividers to add LFMXO5. |
| sysCLOCK PLL | <ul style="list-style-type: none"> Added information on two PLLs for LFMXO5 in sysCLOCK PLL Overview. Added Figure 14.6. PLL Input Pins for LFMXO5. |

Revision 1.8, February 2022

| Section | Change Summary |
|---------------------------------|--------------------------------------------------------------------------------------------|
| Overview of Clocking Components | Changed oscillator output frequency from +/-15% to +/-7% in Section 5.6 . |
| Dynamic Clock Control (DCC) | Added Table 9.2. DCC Component Attribute Definition for DCCEN information. |

Revision 1.7, December 2021

| Section | Change Summary |
|----------------------|--------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------|
| All | <ul style="list-style-type: none"> Minor adjustments in formatting across the document. Changed document title from sysCLOCK PLL Design and Usage Guide for Nexus Platform to sysCLOCK PLL Design and User Guide for Nexus Platform. |
| Introduction | Updated Table 1.1 . |
| Nexus Top-Level View | Added Figure 3.3 . |
| sysCLOCK PLL | Updated sysCLOCK PLL Overview content to add info on two PLLs and added Figure 14.6 . |

Revision 1.6, September 2021

| Section | Change Summary |
|----------------------|-------------------------------------------------------------------|
| Nexus Top-Level View | Changed '4' to '6' for PLL blocks in Figure 3.2 . |

Revision 1.5, July 2021

| Section | Change Summary |
|-------------------------------|---------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------|
| Acronyms in This Table | Added LMMI definition. |
| Dynamic Clock Control | Changed GPLL to PLL instance. |
| General Routing for Clocks | Added entry for PROHIBIT PRIMARY as a workaround for non-PCLK located clock nets. |
| sysCLOCK PLL | <ul style="list-style-type: none"> Changed PLLCLK to PLLCK in Dedicated PLL Inputs, including Figure 14.2 to Figure 14.5. Updated Table 14.3 to correct the values in PLL Output Shifted column. Changed PHASESEL[2:0]=3'b000 to PHASESEL[2:0]=3'b001 in VCO Phase Shift. Added Total Phase Shift, Fractional-N Synthesis Operation, and Spread Spectrum Clock Generation sections. |
| Appendix C.PLL LMMI Operation | Added this section. |

Revision 1.4, June 2021

| Section | Change Summary |
|----------------------------------------------------|----------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------|
| All | <ul style="list-style-type: none"> Minor formatting across the document. Added CertusPro-NX support across the document |
| Introduction | Updated section content, including Table 1.1. |
| Clock/Control Distribution Network | Updated section content to add CertusPro-NX. |
| Nexus Top-Level View | Updated section content, including Figure 3.2. |
| Clocking Architecture Overview | Updated section content to add CertusPro-NX. |
| Overview of Clocking Components | <ul style="list-style-type: none"> Changed section title from Overview of Other Clocking Elements to Overview of Clocking Components. Changed PCIKDIV to PCLKDIV in Primary Clock Divider (PCLKDIV). Updated content to add CertusPro-NX in Dynamic Clock Select (DCS). |
| Primary Clocks | <ul style="list-style-type: none"> Updated content, including adding two bullet points in Primary Clock Sources. Updated content in Primary Clock Routing, including Figure 6.1 and Figure 6.2. |
| Primary Clock Divider | <ul style="list-style-type: none"> Updated Table 7.2. Changed PLKDIVF to PCLKDIV in PCLKDIV Usage in Verilog. |
| Dynamic Clock Select | Updated section content, including Figure 8.1 and Figure 8.2. |
| Dynamic Clock Control | Updated section content, including adding DCC Usage in Verilog. |
| Internal Oscillator | Updated section content to include CertusPro-NX. |
| Edge Clocks | <ul style="list-style-type: none"> Added bullet point for Bottom PLL Outputs. Updated Table 11.2. Updated codes in ECLKDIV Usage in VHDL and ECLKDIV Usage in Verilog. |
| Edge Clock Synchronization | Updated codes in ECLKSYNC Usage in Verilog. |
| sysCLOCK PLL | <ul style="list-style-type: none"> Updated section content, including Figure 14.4 and Figure 14.5. Updated Table 14.1, Table 14.3, Table 14.6, and Table 14.7. Updated equations in VCO Phase Shift and Divider Phase Shift. |
| Appendix A. Primary Clock Sources and Distribution | Added Figure A.3 and Figure A.4. |

Revision 1.3, November 2020

| Section | Change Summary |
|---------------------|----------------------------------------------------------------------------------------------------------------------------------------------------------------------|
| All | Changed some OSC instances to OSCA. |
| Internal Oscillator | <ul style="list-style-type: none"> Updated Table 10.1 and Table 10.2. Updated codes in OSCA Usage in VHDL and OSCA Usage in Verilog section. |
| sysCLOCK PLL | Updated Table 14.6. |

Revision 1.2, June 2020

| Section | Change Summary |
|----------------------|------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------|
| All | <ul style="list-style-type: none"> Changed document name to sysCLOCK PLL Design and Usage Guide for Nexus Platform. Changed CrossLink-NX to Nexus across the document. |
| Nexus Top-Level View | Updated content. |
| sysCLOCK PLL | <ul style="list-style-type: none"> Updated content to add LFD2NX-17 and LFD2NX-40. Moved PLL Features to this section. Updated Table 14.7. |

Revision 1.1, April 2020

| Section | Change Summary |
|----------------|-----------------------------|
| PLL Features | Updated Figure 7.1 and 7.2. |
| Primary Clocks | Updated Figure 6.1. |
| sysCLOCK PLL | Updated Figure 16.3. |

Revision 1.0, November 2019

| Section | Change Summary |
|---------|-----------------|
| All | Initial release |



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